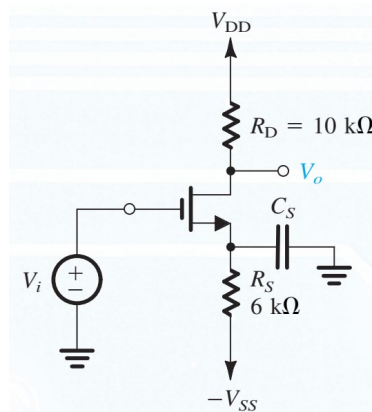
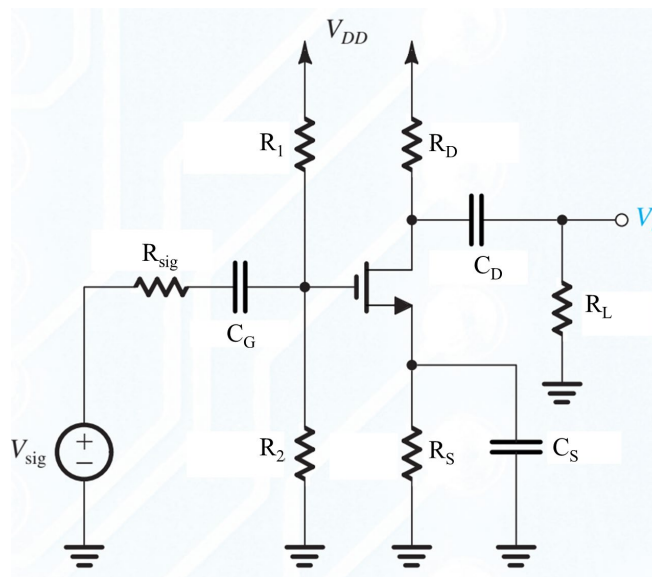


1. The amplifier in the figure below is biased to operate at  $g_m = 1 \text{ mA/V}$ . Neglecting  $r_o$ ,
  - a. Find the mid-band gain.
  - b. Find the full transfer function with  $C_S$  present. Is this circuit low-pass or high-pass?
  - c. Find the value of  $C_S$  that places  $f_L$  at 200 Hz.
  - d. With the value of  $C_S$  above, find  $v_o(t)$ . Given that  $v_i(t) = V_0 + 10 \cos(10^6 \pi t) \text{ mV}$  where  $V_0$  is a constant.



2. The NMOS transistor in the discrete CS amplifier circuit shown below is biased to have  $g_m = 5 \text{ mA/V}$ . Find  $A_M$ ,  $f_{p1}$ ,  $f_{p2}$ ,  $f_{p3}$ , and  $f_L$ . Let  $R_{\text{sig}} = 100 \text{ k}\Omega$ ,  $R_1 = 47 \text{ M}\Omega$ ,  $R_2 = 10 \text{ M}\Omega$ ,  $R_D = 4.7 \text{ k}\Omega$ ,  $R_S = 2 \text{ k}\Omega$ ,  $R_L = 10 \text{ k}\Omega$ ,  $C_G = 0.01 \mu\text{F}$ ,  $C_S = 10 \mu\text{F}$ ,  $C_D = 0.1 \mu\text{F}$



Ignore parasitic capacitances of the transistor and channel-length modulation effect.

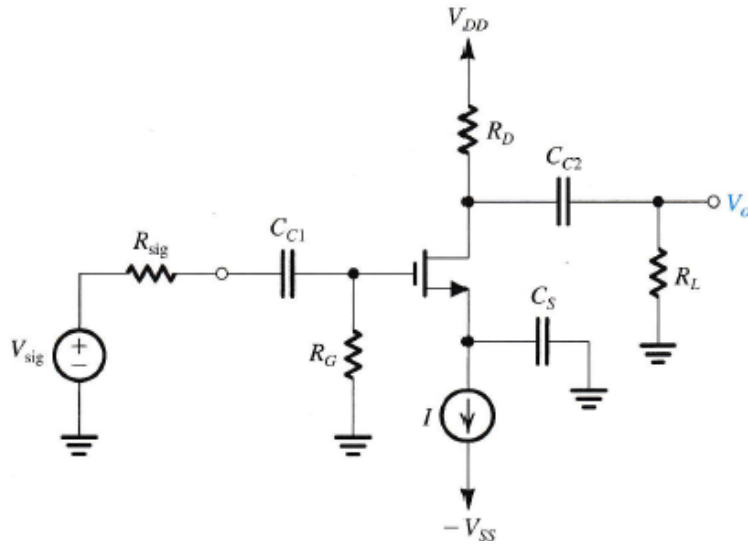
3. A discrete MOSFET common-source amplifier has  $R_G = 1M\Omega$ ,  $g_m = 5mA/V$ ,  $r_o = 100k\Omega$ ,  $R_D = 10k\Omega$ ,  $C_{gs} = 2pF$ , and  $C_{gd} = 0.4pF$ . The amplifier is fed from a voltage source with an internal resistance of  $500k\Omega$  and is connected to a  $10k\Omega$  load. Find:

a) The overall mid-band gain  $A_M$

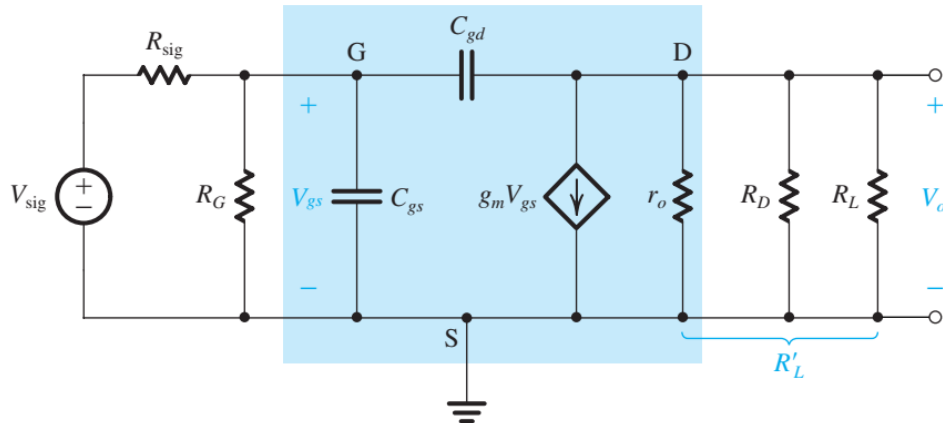
b) The upper 3-dB frequency  $f_H$

Refer to Figs a and b for the complete setup and the small signal circuit at high frequency. Note: at high and mid-band frequency, coupling capacitor  $C_{C1}$ ,  $C_{C2}$ ,  $C_S$  are shorted.

**Hint:** In part b), find  $f_H$  using Miller's theorem then apply open-circuit time constant approach.



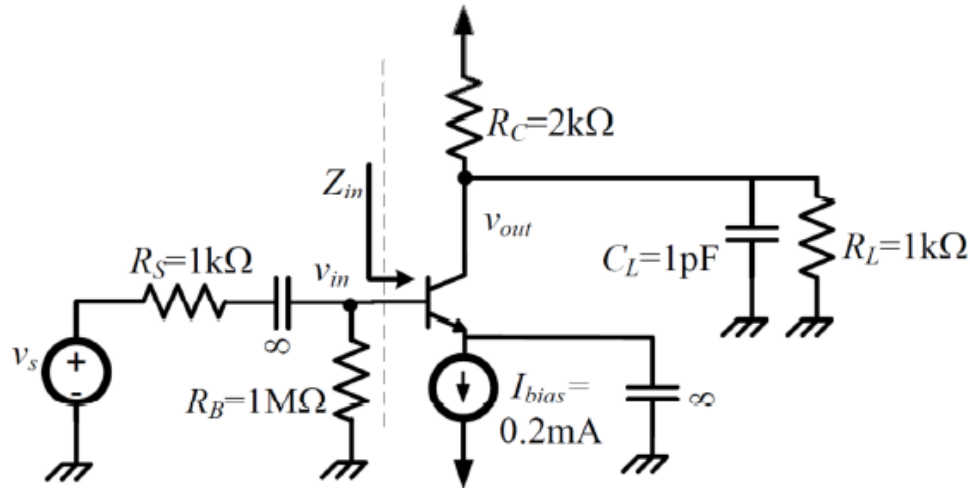
(a)



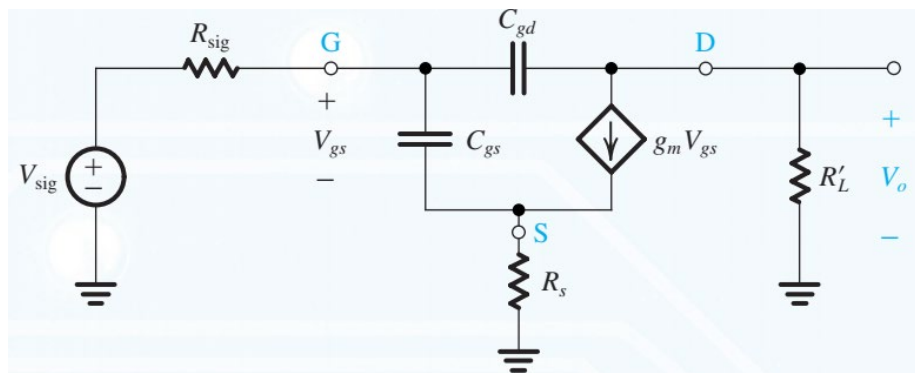
(b)

4. Consider the common-emitter amplifier in the following figure, with  $\beta = 100$ ,  $V_A = 100V$ ,  $C_\pi = 25fF$ ,  $C_\mu = 10fF$ .

- Draw the small-signal model of this circuit. Apply Miller's theorem to split  $C_\mu$  to input and output nodes. Calculate the time constants at the input and output nodes,  $\tau_{in}$  and  $\tau_{out}$ .
- Based on the time constants from part a), calculate the input and output pole frequencies,  $f_{in}$  and  $f_{out}$ . What is the dominant pole of this amplifier?



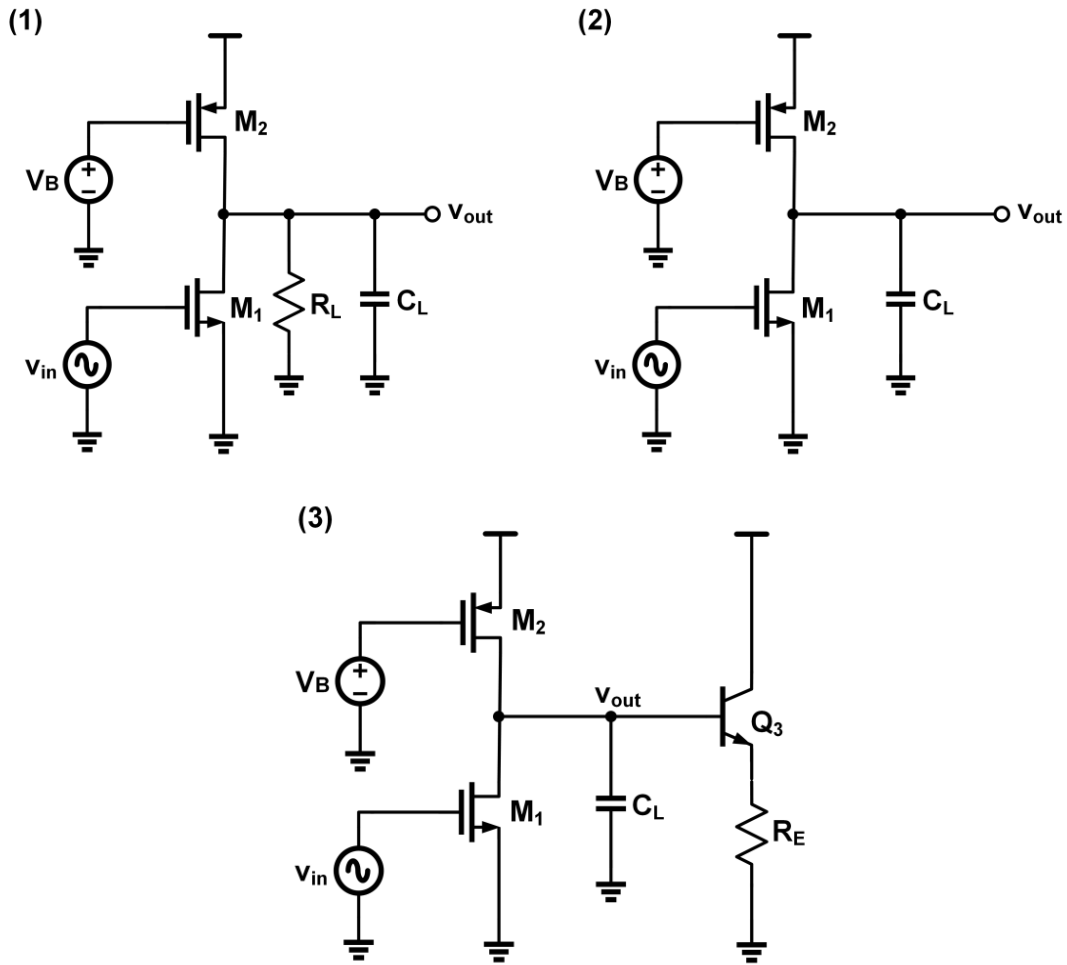
5. The figure below shows the high-frequency equivalent circuit of a CS amplifier with a resistance  $R_s$ , connected to S. The purpose of this problem is to show that the value of  $R_s$ , can be used to control the gain and bandwidth of the amplifier, specifically to allow the designer to trade gain for increased bandwidth.



- Derive an expression for the low-frequency voltage gain (i.e. set  $C_{gs}$  and  $C_{gd}$  to zero).
- To be able to determine  $\omega_H$  using the open-circuit time-constants method, derive expressions for  $R_{gs}$  and  $R_{gd}$  (equivalent resistance seen by  $C_{gs}$  and  $C_{gd}$ , respectively).

- c) Let  $R_{sig} = 100k\Omega$  ,  $g_m = 4mA/V$  ,  $R'_L = 5k\Omega$  , and  $C_{gs} = C_{gd} = 1pF$  . Use the expressions found in a) and b) to determine the low-frequency gain and the 3-dB frequency  $f_H$  for three cases:  $R_s = 0\Omega$ ,  $100\Omega$ , and  $250\Omega$ . In each case, also evaluate the gain-bandwidth product.

6. Determine -3dB bandwidth of the circuits shown below. Assume MOS transistors in saturation and BJTs in forward active region with  $r_{ds} = \infty$ ,  $r_o = \infty$ . Ignore intrinsic capacitances.



7. Approximate transfer function for the circuits below. Assume MOS transistors operate in saturation with  $r_{ds} = \infty$ , and BJTs in forward active region with  $r_o = \infty$

