

ECE 342

Electronic Circuits

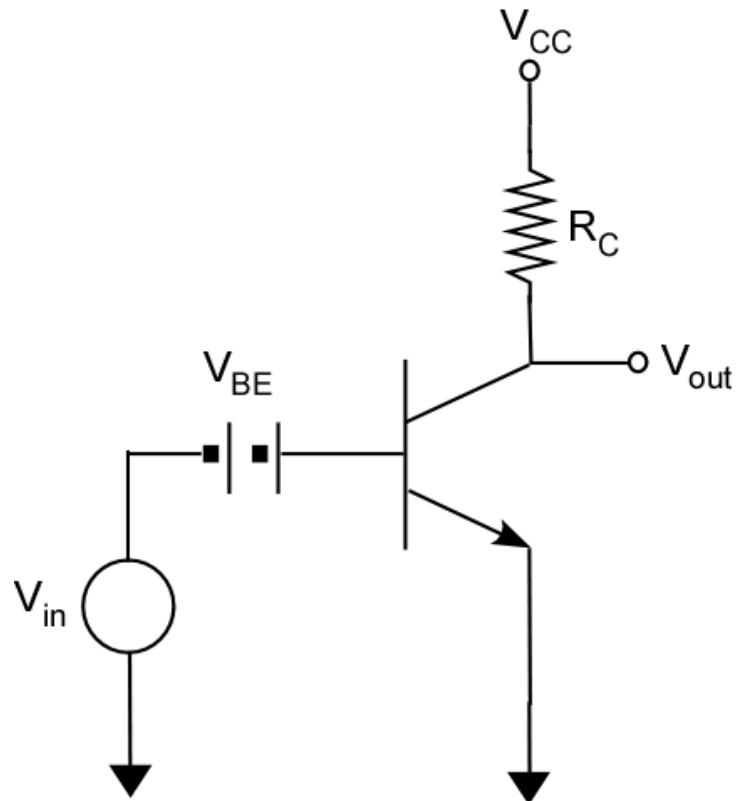
Lecture 17

Common Emitter - Part 1

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Common Emitter Configuration

The emitter current I_E can be approximated as:



$$I_E \approx I_S e^{V_{BE}/V_T}$$

An incremental conductance g_e can be defined as

$$g_e \equiv \frac{\partial I_E}{\partial V_{BE}} = \frac{I_S}{V_T} e^{V_{BE}/V_T} = \frac{I_E}{V_T}$$

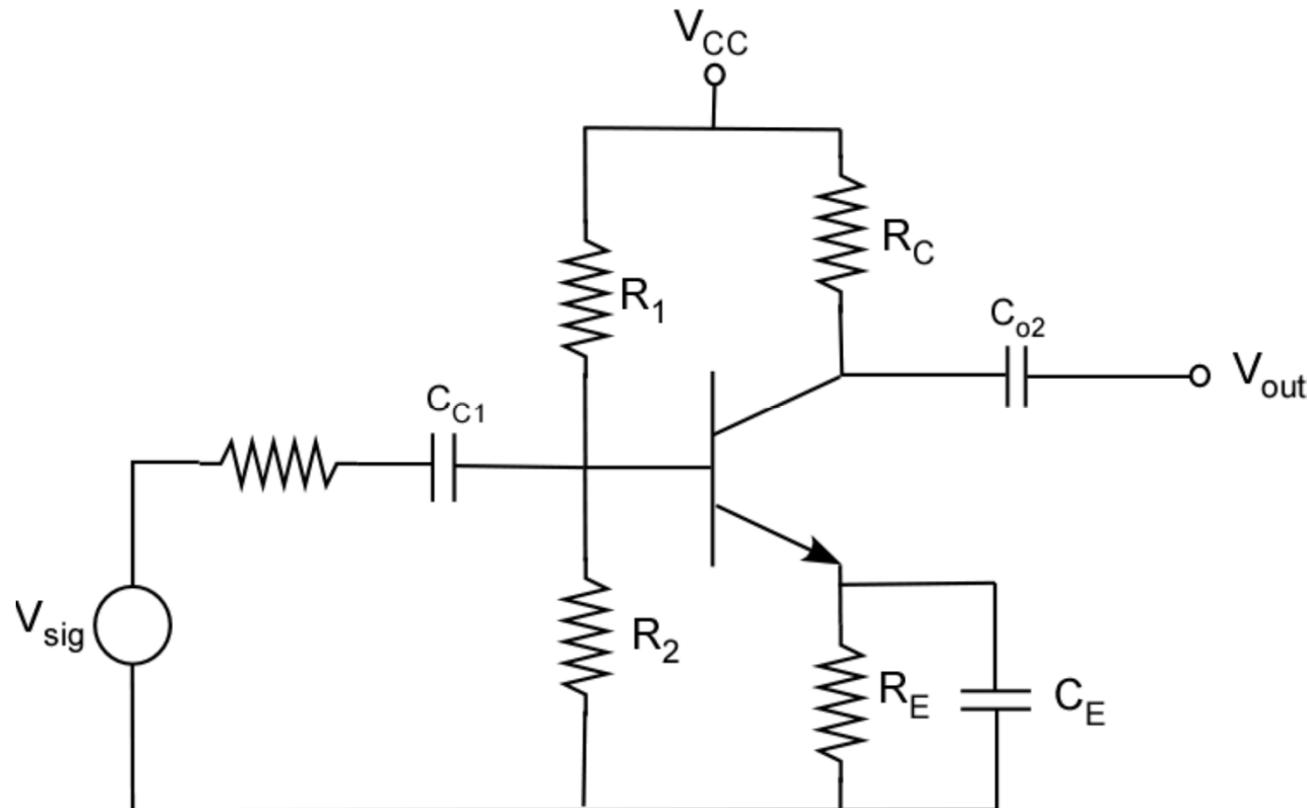
From which we get:

$$r_e = \frac{1}{g_e} = \frac{V_T}{I_E}$$

Emitter resistance

$$\text{Usually, } V_T = 26 \text{ mV} \Rightarrow r_e = \frac{26 \text{ mV}}{I_E}$$

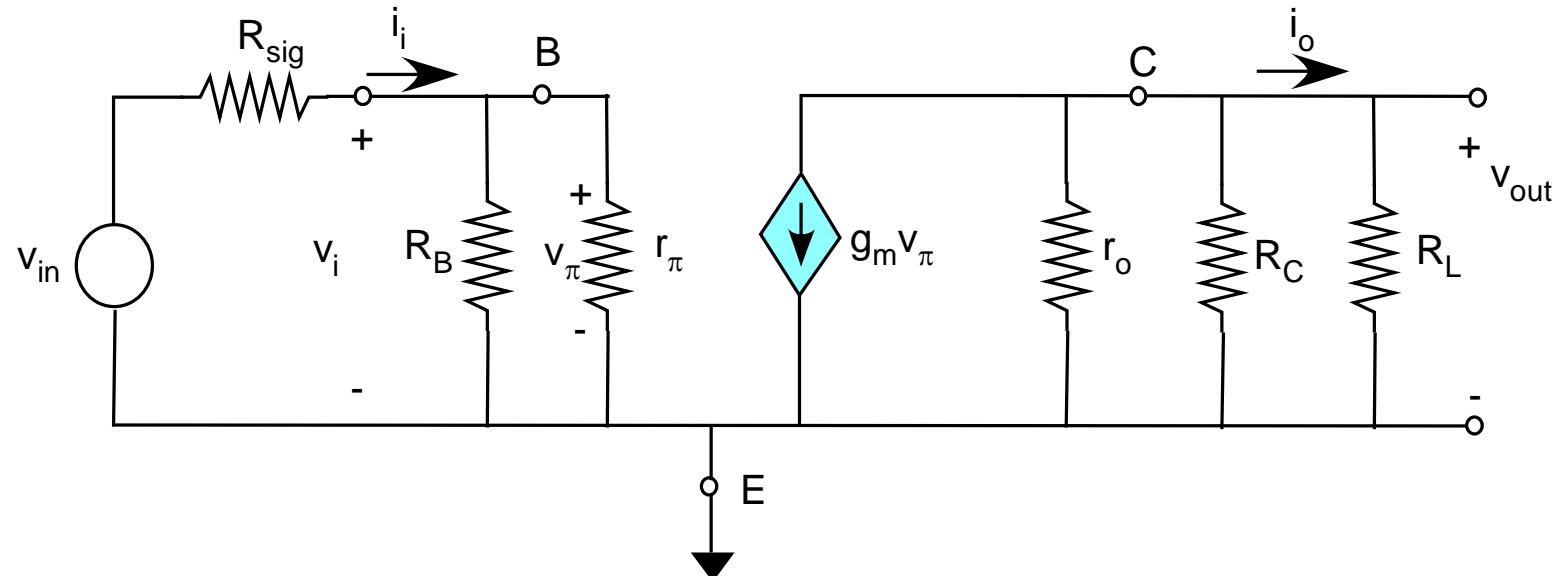
Common Emitter (CE) Amplifier



Bias: Choose R_1 & R_2 to set $V_B \rightarrow V_E$ is then set. Choose R_E to set $I_E \sim I_C$. Quiescent point of V_{out} will be determined by R_C . Emitter is an AC short.

Incremental Model for CE Amplifier

Hybrid- π model (ignoring r_x)



$$R_B = R_1 \parallel R_2$$

$$R_{in} = \frac{v_i}{i_i} = R_B \parallel r_\pi$$

Sometimes $R_B \gg r_\pi$ and $R_{in} \simeq r_\pi$

CE Amplifier

$$v_i = \frac{v_{sig} R_{in}}{R_{in} + R_{sig}} = \frac{v_{sig} R_B \parallel r_\pi}{(R_B \parallel r_\pi) + R_{sig}}$$

and if $R_B \gg r_\pi$, $v_i \simeq \frac{v_{sig} r_\pi}{r_\pi + R_{sig}}$

$$v_o = -g_m v_{sig} \frac{R_B \parallel r_\pi (r_o \parallel R_C \parallel R_L)}{(R_B \parallel r_\pi) + R_{sig}}$$

$$v_o = -g_m v_\pi (r_o \parallel R_C \parallel R_L)$$

$$v_i = v_\pi$$

$$A_v = \frac{v_o}{v_i} = -g_m (r_o \parallel R_C \parallel R_L)$$

gain from base to collector

CE Amplifier

Open-circuit voltage gain:

$$A_{vo} = -g_m (r_o \parallel R_C)$$

In most cases $r_o \gg R_C \Rightarrow A_{vo} = -g_m R_C$

$$G_v = -\frac{(R_B \parallel r_\pi)}{(R_B \parallel r_\pi) + R_{sig}} g_m (r_o \parallel R_C \parallel R_L)$$

and for the case where $R_B \gg r_\pi$

$$G_v = -\frac{\beta (r_o \parallel R_C \parallel R_L)}{r_\pi + R_{sig}}$$

CE Amplifier

Output Impedance

$$R_{out} = R_C \parallel r_o$$

If $r_o \gg R_C$, $R_{out} \simeq R_C$

from which $A_v = A_{vo} \left(\frac{R_L}{R_L + R_o} \right)$

It can be seen that if $R_{sig} \gg r_\pi$, the gain will be highly dependent on β . This is not good because of β variations

If $R_{sig} \ll r_\pi$, $G_v \simeq -g_m (R_C \parallel R_L \parallel r_o)$