

ECE 451

Coupled Lines

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Crosstalk Noise

Signal Integrity

Crosstalk

Dispersion

Attenuation

Reflection

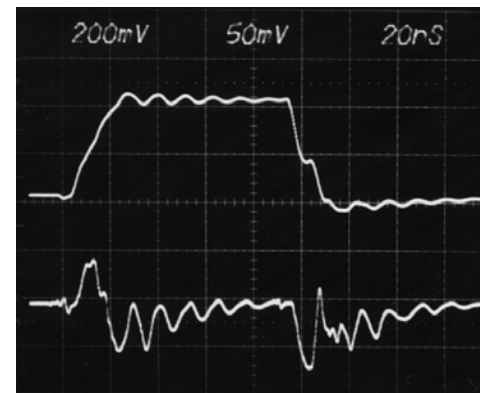
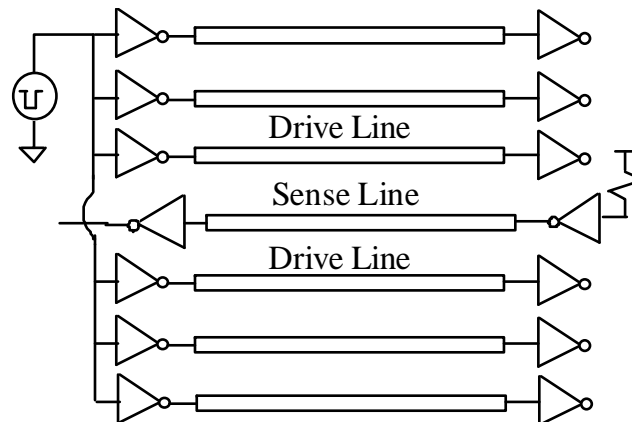
Distortion

Loss

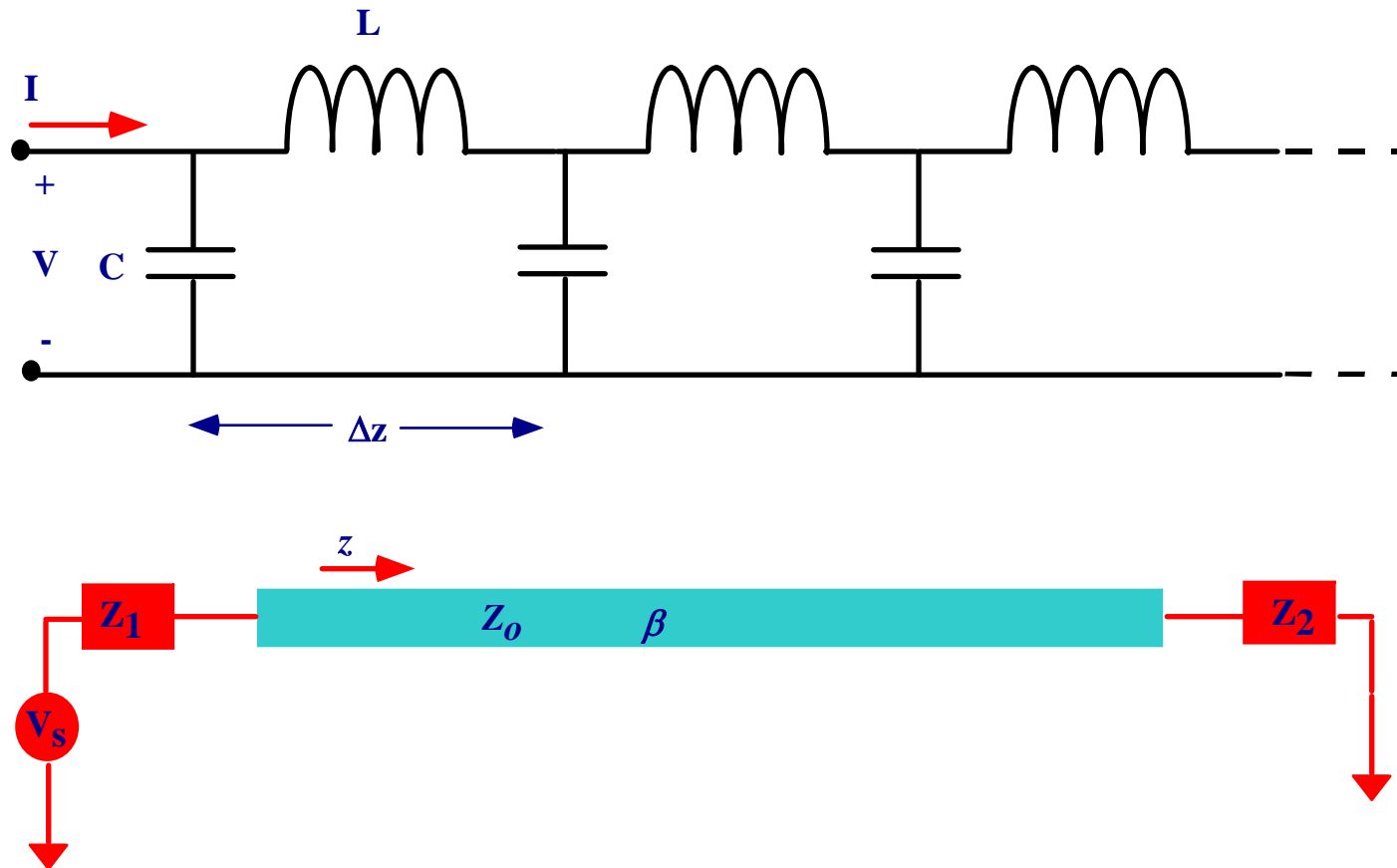
Delta I Noise

Ground Bounce

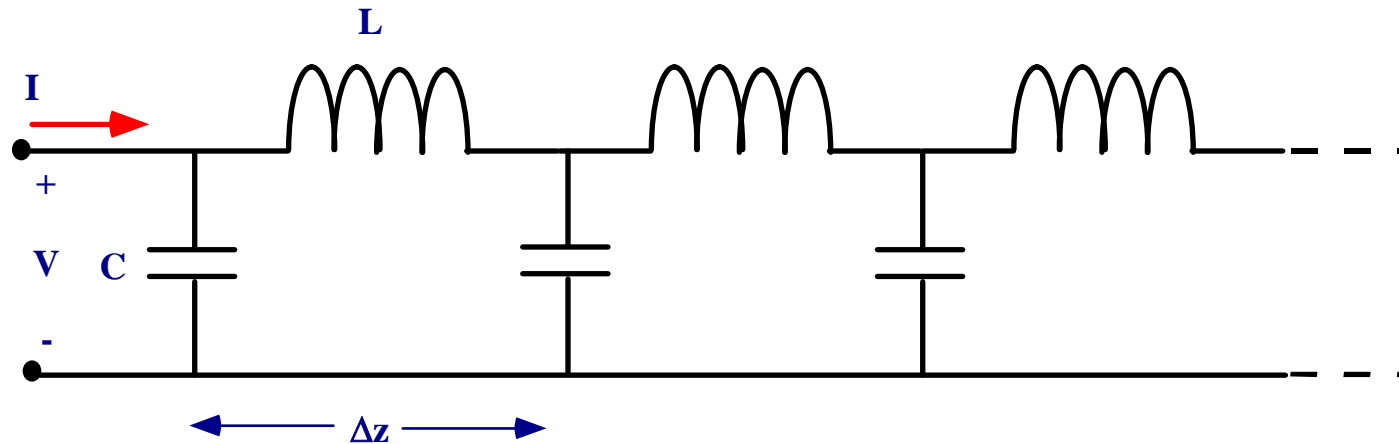
Radiation



TEM PROPAGATION



Telegrapher's Equations



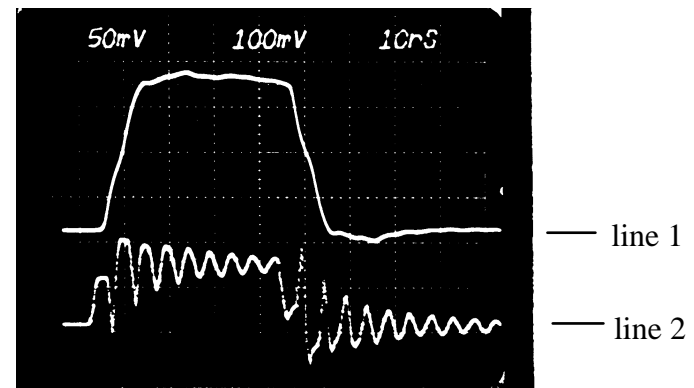
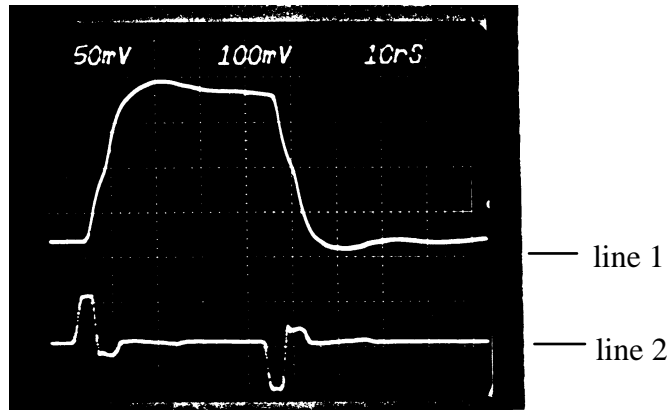
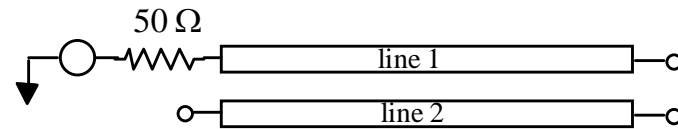
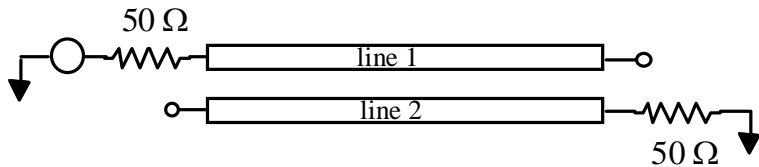
$$-\frac{\partial V}{\partial z} = L \frac{\partial I}{\partial t}$$

$$-\frac{\partial I}{\partial z} = C \frac{\partial V}{\partial t}$$

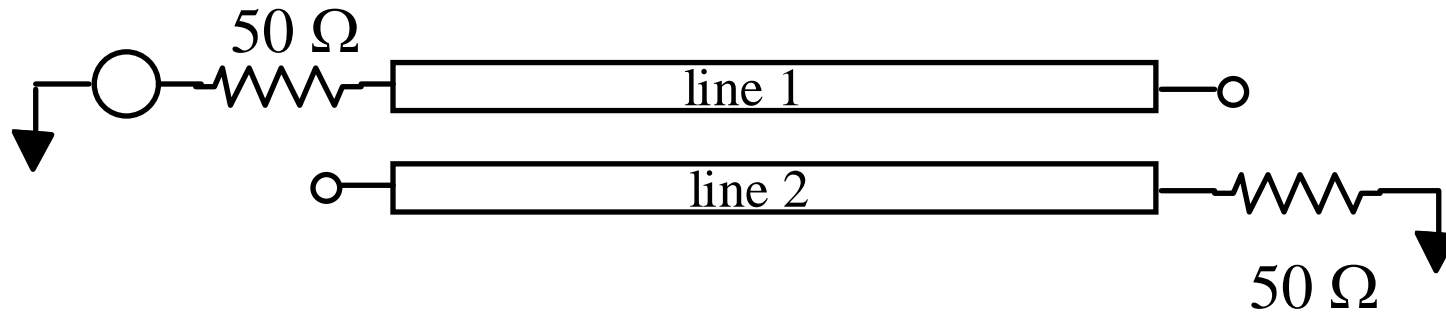
L: Inductance per unit length.

C: Capacitance per unit length.

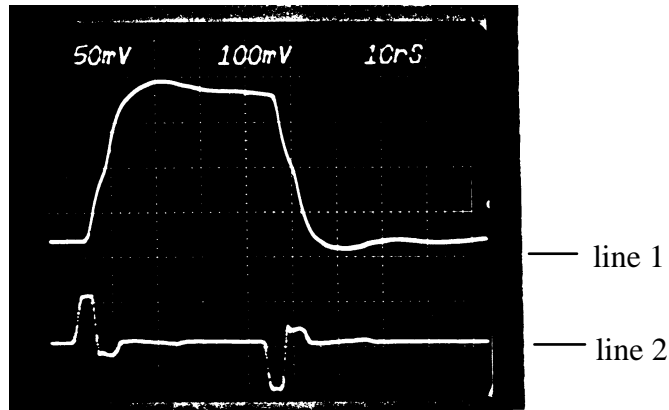
Crosstalk noise depends on termination



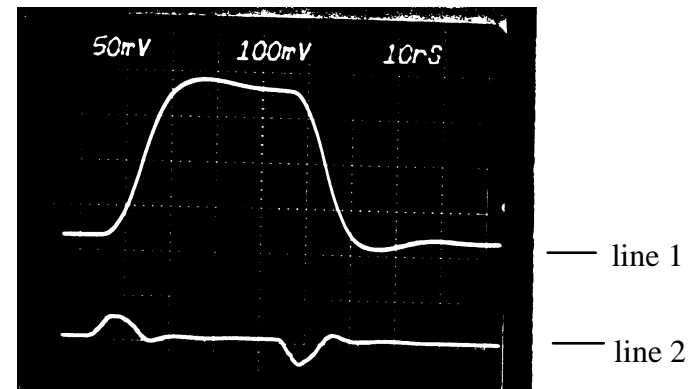
Crosstalk depends on signal rise time



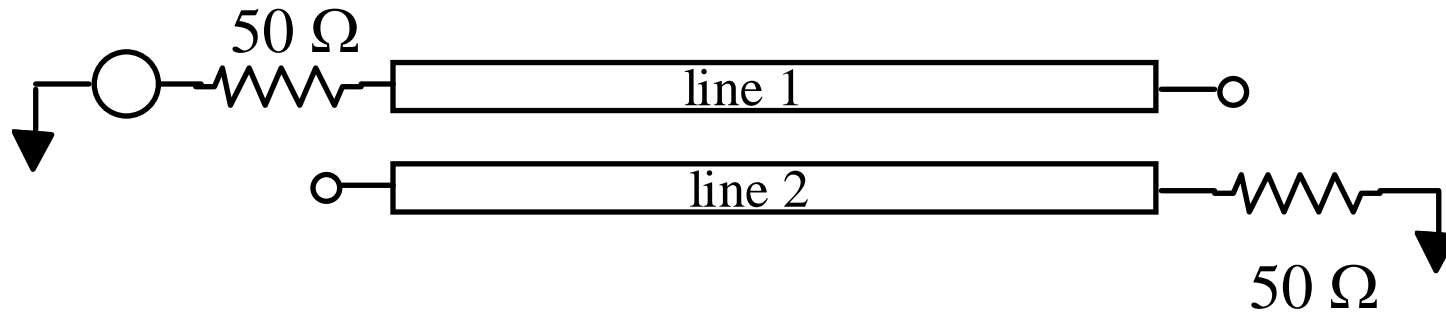
$t_r = 1\ \text{ns}$



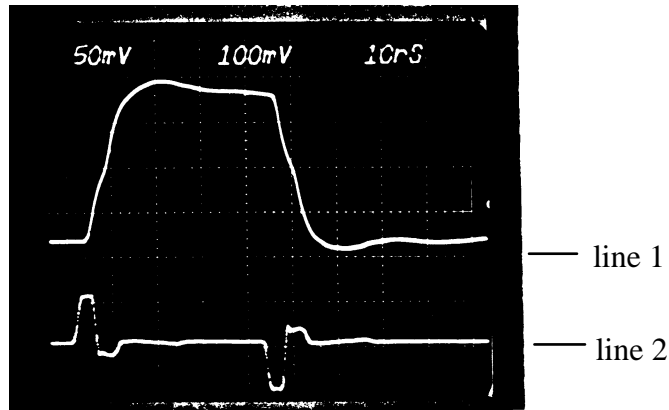
$t_r = 7\ \text{ns}$



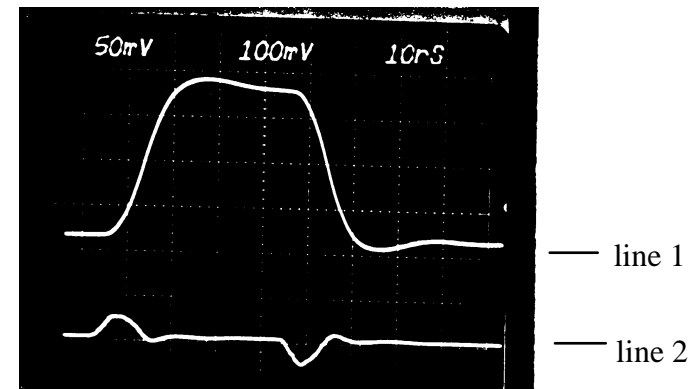
Crosstalk depends on signal rise time



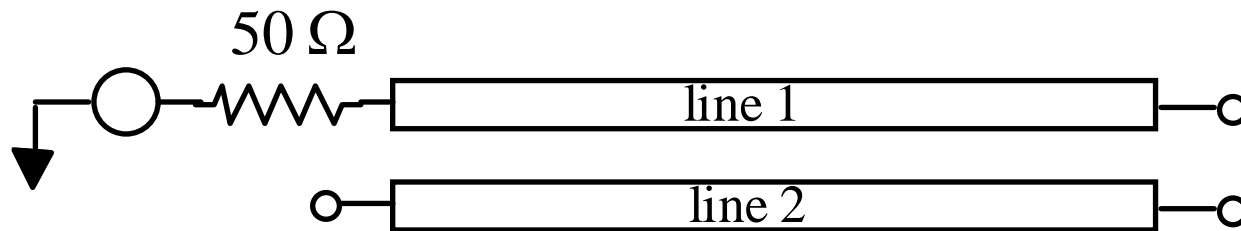
$t_r = 1\ \text{ns}$



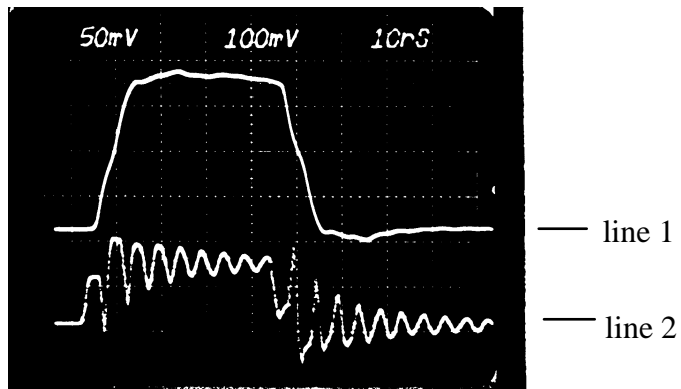
$t_r = 7\ \text{ns}$



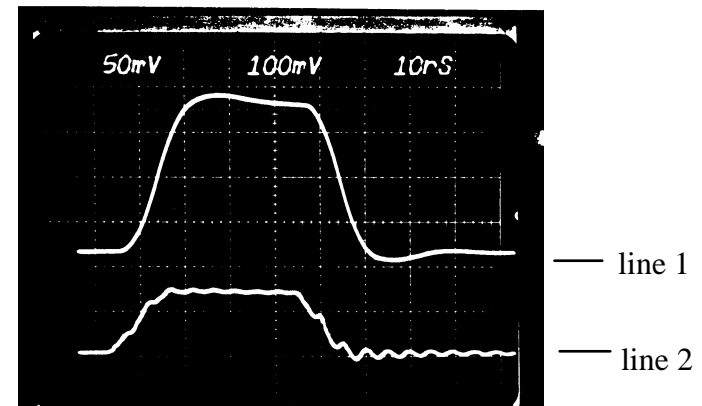
Crosstalk depends on signal rise time



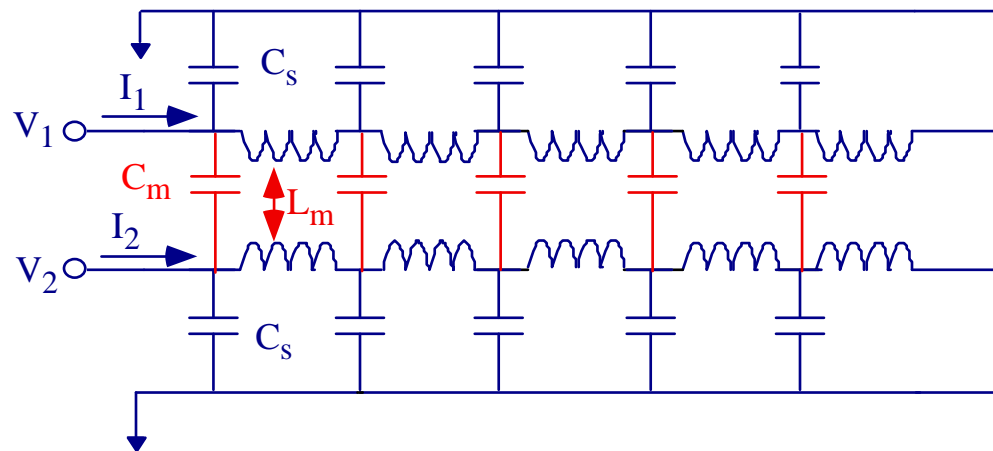
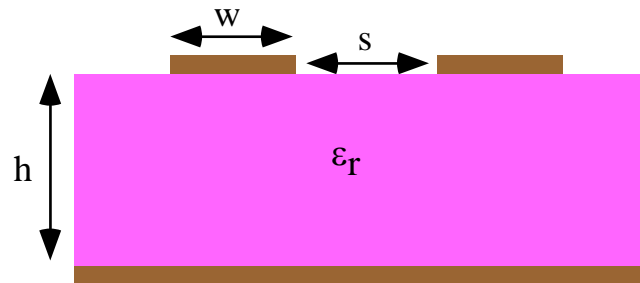
$t_r = 1\ \text{ns}$



$t_r = 7\ \text{ns}$



Coupled Transmission Lines



Telegraphers Equations for Coupled Transmission Lines

Maxwellian Form

$$-\frac{\partial V_1}{\partial z} = L_{11} \frac{\partial I_1}{\partial t} + L_{12} \frac{\partial I_2}{\partial t}$$

$$-\frac{\partial V_2}{\partial z} = L_{21} \frac{\partial I_1}{\partial t} + L_{22} \frac{\partial I_2}{\partial t}$$

$$-\frac{\partial I_1}{\partial z} = C_{11} \frac{\partial V_1}{\partial t} + C_{12} \frac{\partial V_2}{\partial t}$$

$$-\frac{\partial I_2}{\partial z} = C_{21} \frac{\partial V_1}{\partial t} + C_{22} \frac{\partial V_2}{\partial t}$$

Telegraphers Equations for Coupled Transmission Lines

Physical form

$$-\frac{\partial V_1}{\partial z} = L_s \frac{\partial I_1}{\partial t} + L_m \frac{\partial I_2}{\partial t}$$

$$-\frac{\partial V_2}{\partial z} = L_m \frac{\partial I_1}{\partial t} + L_s \frac{\partial I_2}{\partial t}$$

$$-\frac{\partial I_1}{\partial z} = C_s \frac{\partial V_1}{\partial t} + C_m \frac{\partial V_1}{\partial t} - C_m \frac{\partial V_2}{\partial t}$$

$$-\frac{\partial I_2}{\partial z} = -C_m \frac{\partial V_1}{\partial t} + C_m \frac{\partial V_2}{\partial t} + C_s \frac{\partial V_2}{\partial t}$$

Relations Between Physical and Maxwellian Parameters (symmetric lines)

$$L_{11} = L_{22} = L_s$$

$$L_{12} = L_{21} = L_m$$

$$C_{11} = C_{22} = C_s + C_m$$

$$C_{12} = C_{21} = - C_m$$

Even Mode

$$-\frac{\partial V_e}{\partial z} = (L_{11} + L_{12}) \frac{\partial I_e}{\partial t}$$

$$-\frac{\partial I_e}{\partial z} = (C_{11} + C_{12}) \frac{\partial V_e}{\partial t}$$

**Add voltage
and current
equations**

V_e : Even mode voltage $V_e = \frac{1}{2}(V_1 + V_2)$

I_e : Even mode current $I_e = \frac{1}{2}(I_1 + I_2)$

$$Z_e = \sqrt{\frac{L_{11} + L_{12}}{C_{11} + C_{12}}} = \sqrt{\frac{L_s + L_m}{C_s}}$$

Impedance

$$v_e = \frac{1}{\sqrt{(L_{11} + L_{12})(C_{11} + C_{12})}} = \frac{1}{\sqrt{(L_s + L_m)C_s}}$$

velocity

Odd Mode

$$-\frac{\partial V_d}{\partial z} = (L_{11} - L_{12}) \frac{\partial I_d}{\partial t}$$

$$-\frac{\partial I_d}{\partial z} = (C_{11} - C_{12}) \frac{\partial V_d}{\partial t}$$

**Subtract voltage
and current
equations**

V_d : Odd mode voltage

$$V_d = \frac{1}{2}(V_1 - V_2)$$

I_d : Odd mode current

$$I_d = \frac{1}{2}(I_1 - I_2)$$

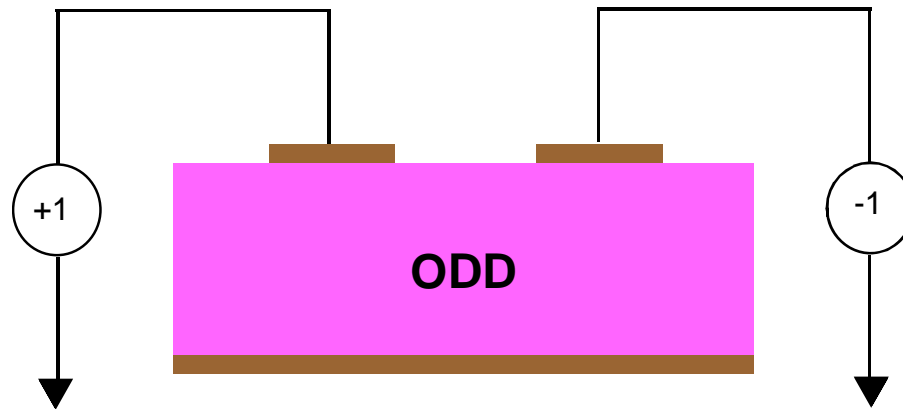
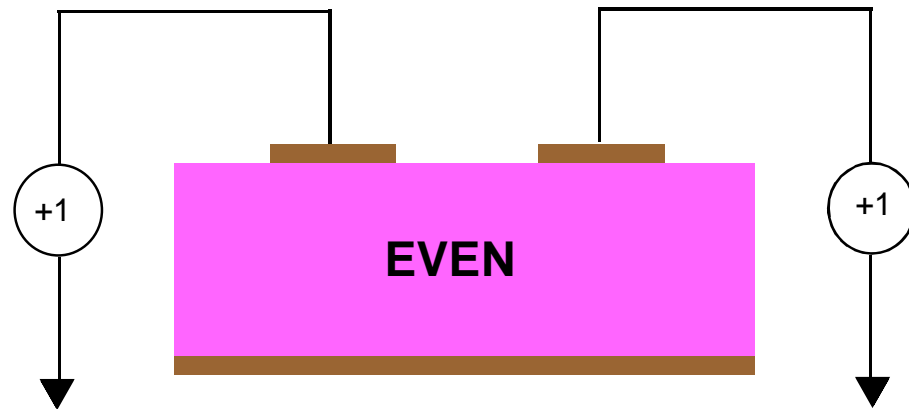
$$Z_d = \sqrt{\frac{L_{11} - L_{12}}{C_{11} - C_{12}}} = \sqrt{\frac{L_s - L_m}{C_s + 2C_m}}$$

Impedance

$$v_d = \frac{1}{\sqrt{(L_{11} - L_{12})(C_{11} - C_{12})}} = \frac{1}{\sqrt{(L_s - L_m)(C_s + 2C_m)}}$$

velocity

Mode Excitation



PHYSICAL SIGNIFICANCE OF EVEN- AND ODD-MODE IMPEDANCES

- * Z_e and Z_d are the wave resistance seen by the even and odd mode travelling signals respectively.
- * The impedance of each line is no longer described by a single characteristic impedance; instead, we have

$$V_1 = Z_{11} I_1 + Z_{12} I_2$$

$$V_2 = Z_{21} I_1 + Z_{22} I_2$$

Definitions

Even-Mode Impedance: Z_e

Impedance seen by wave propagating through the coupled-line system when excitation is symmetric (1, 1).

Odd-Mode Impedance: Z_d

Impedance seen by wave propagating through the coupled-line system when excitation is anti-symmetric (1, -1).

Common-Mode Impedance: $Z_c = 0.5Z_e$

Impedance seen by a pair of line and a common return by a common signal.

Differential Impedance: $Z_{\text{diff}} = 2Z_d$

Impedance seen across a pair of lines by differential mode signal.

EVEN AND ODD-MODE IMPEDANCES

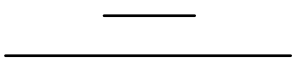
Z_{11}, Z_{22} : Self Impedances

Z_{12}, Z_{21} : Mutual Impedances

For symmetrical lines,

$$\mathbf{Z_{11} = Z_{22} \text{ and } Z_{12} = Z_{21}}$$

EXAMPLE (Microstrip)


$$\epsilon_r = 4.3$$
$$Z_s = 56.4 \Omega$$

Single Line

Dielectric height = 6 mils

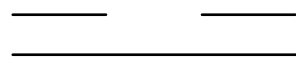
Width = 8 mils

Coupled Lines

Height = 6 mils

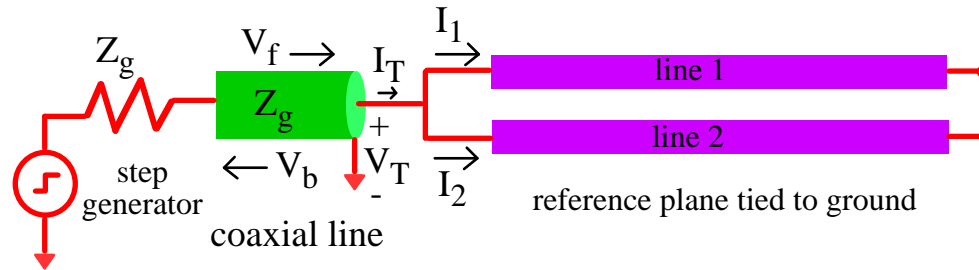
Width = 8 mils

Spacing = 12 mils


$$\epsilon_r = 4.3$$

$$Z_e = 68.1 \Omega \quad Z_d = 40.8 \Omega$$
$$Z_{11} = 54.4 \Omega \quad Z_{12} = 13.6 \Omega$$

Even Mode

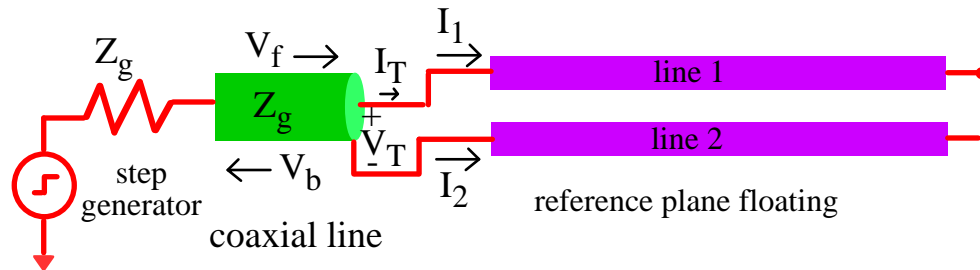


$$I_{tdr} = \left[\frac{a_e(t,0)}{Z_e} + \frac{a_d(t,0)}{Z_d} \right] + \left[\frac{a_e(t,0)}{Z_e} - \frac{a_d(t,0)}{Z_d} \right]$$

$$V_{tdr} = a_e(t,0) - a_d(t,0) \quad a_d(t,0) = 0$$

$$\frac{V_{tdr}}{I_{tdr}} = \frac{Z_e}{2} \quad Z_e = 2 \left(\frac{1 + \rho_e}{1 - \rho_e} \right) Z_g \quad v_e = \frac{2l}{\tau_e}$$

Odd Mode



$$V_{tdr} = a_e(t,0) + a_d(t,0) - [a_e(t,0) - a_d(t,0)] = V_f + V_b$$

$$I_{tdr} = \left[\frac{a_e(t,0)}{Z_e} + \frac{a_d(t,0)}{Z_d} \right] \quad I_{tdr} = - \left[\frac{a_e(t,0)}{Z_e} - \frac{a_d(t,0)}{Z_d} \right]$$

$$a_e(t,0) = 0, \quad \frac{V_{tdr}}{I_{tdr}} = 2Z_d$$

$$Z_d = \frac{1}{2} \left(\frac{1 + \rho_d}{1 - \rho_d} \right) Z_g, \quad v_d = \frac{2l}{\tau_d}$$

EXTRACT INDUCTANCE AND CAPACITANCE COEFFICIENTS

$$L_s = \frac{1}{2} \left[\frac{Z_e}{v_e} + \frac{Z_d}{v_d} \right]$$

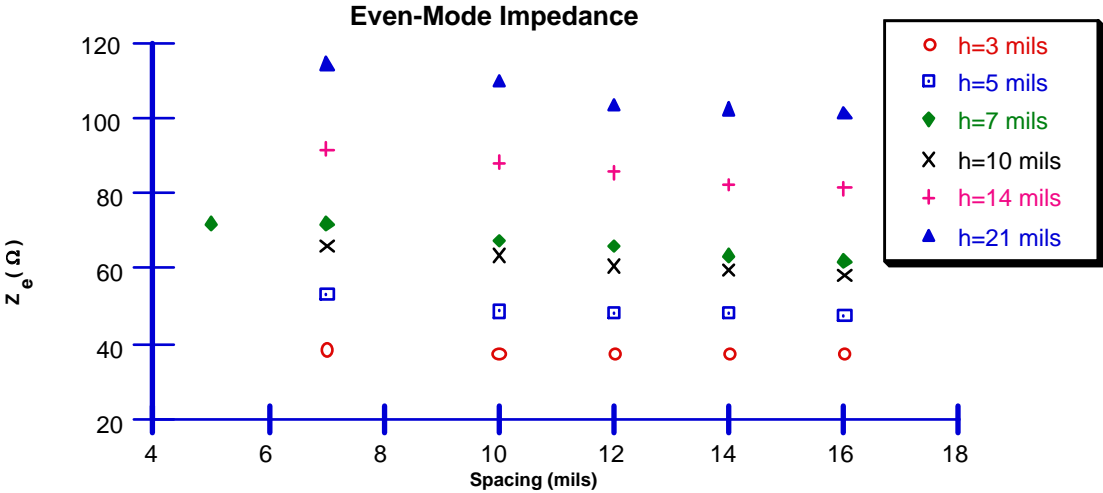
$$C_s = \frac{1}{Z_e v_e}$$

$$L_m = \frac{1}{2} \left[\frac{Z_e}{v_e} - \frac{Z_d}{v_d} \right]$$

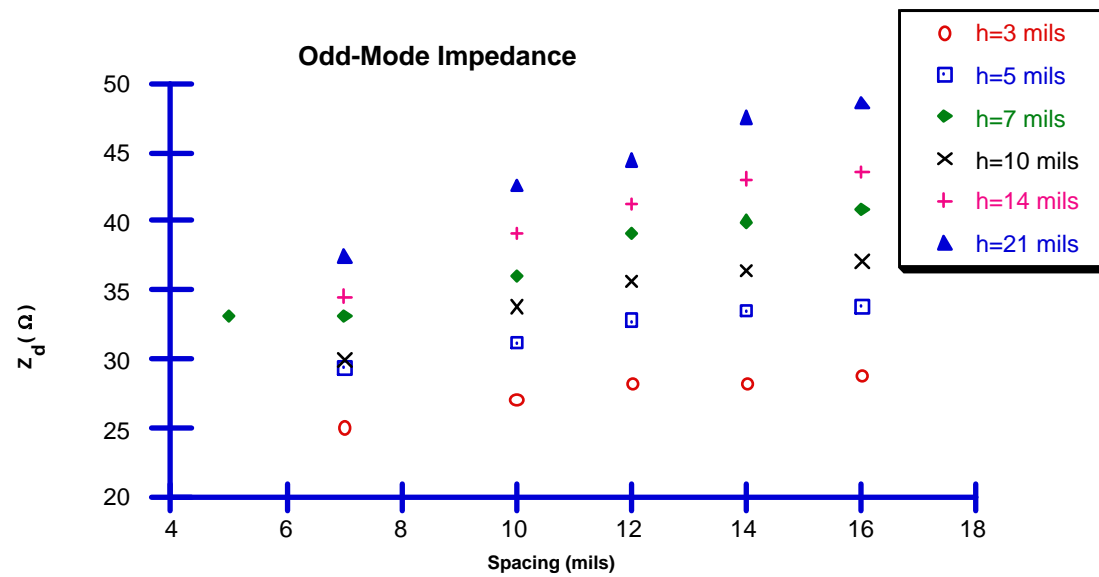
$$C_m = \frac{1}{2} \left[\frac{1}{Z_e v_e} - \frac{1}{Z_d v_d} \right]$$

$$Z_d < Z_s < Z_e$$

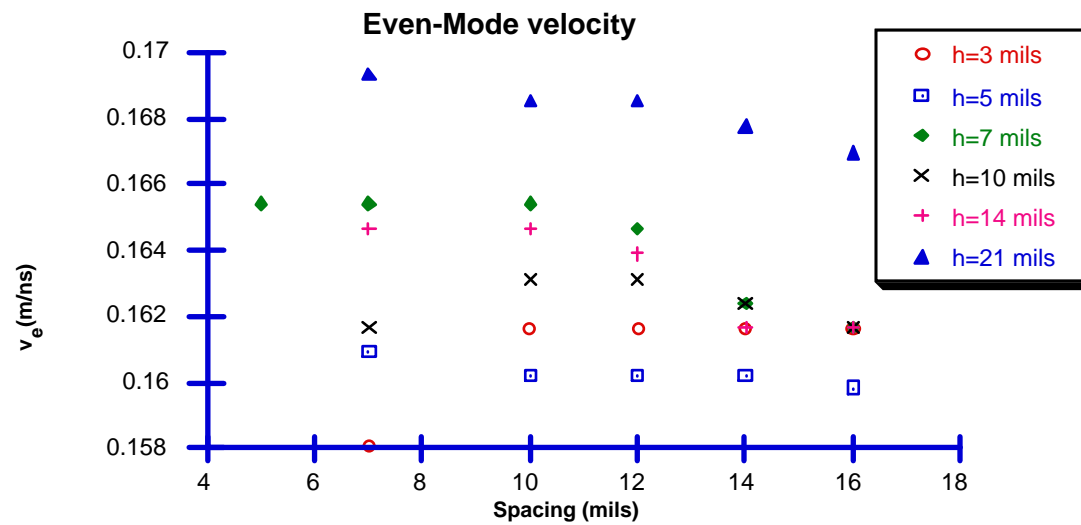
Measured even-mode impedance



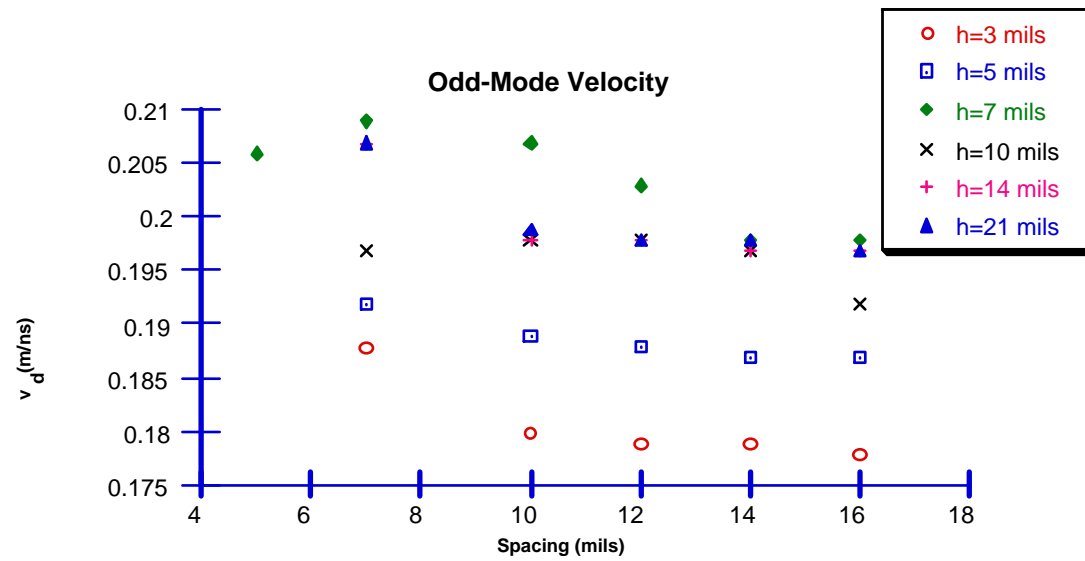
Measured odd-mode impedance



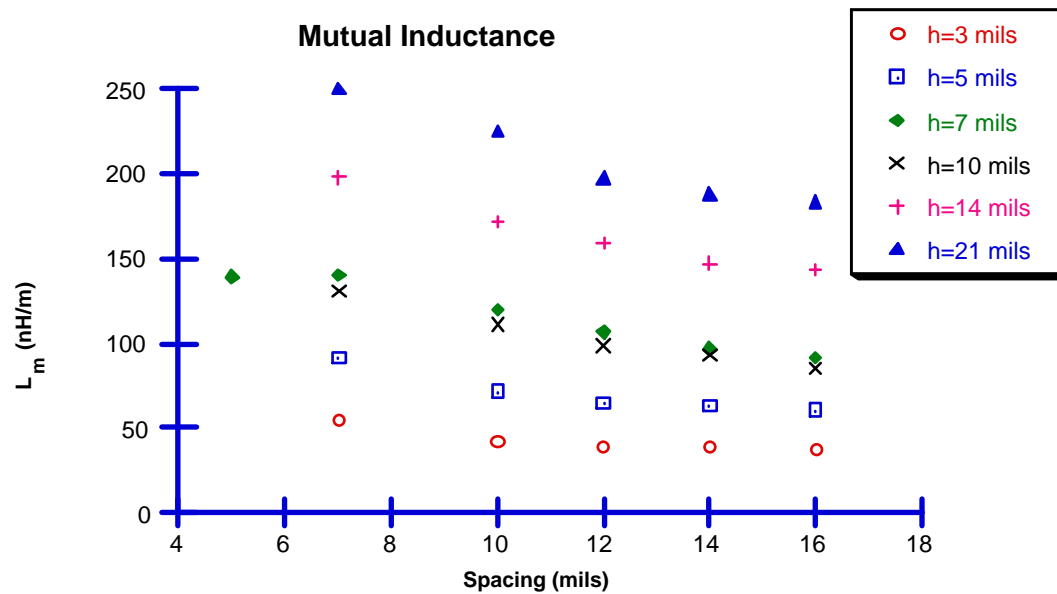
Measured even-mode velocity



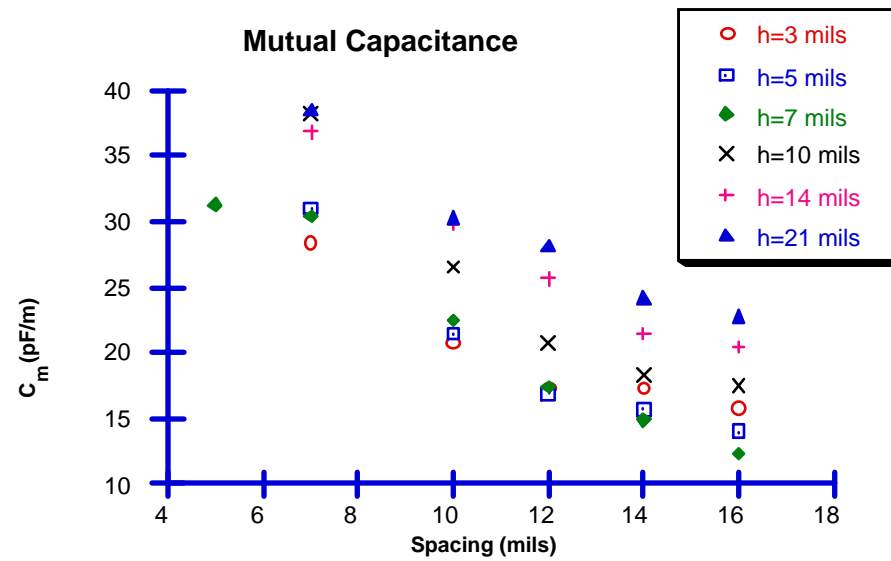
Measured odd-mode velocity



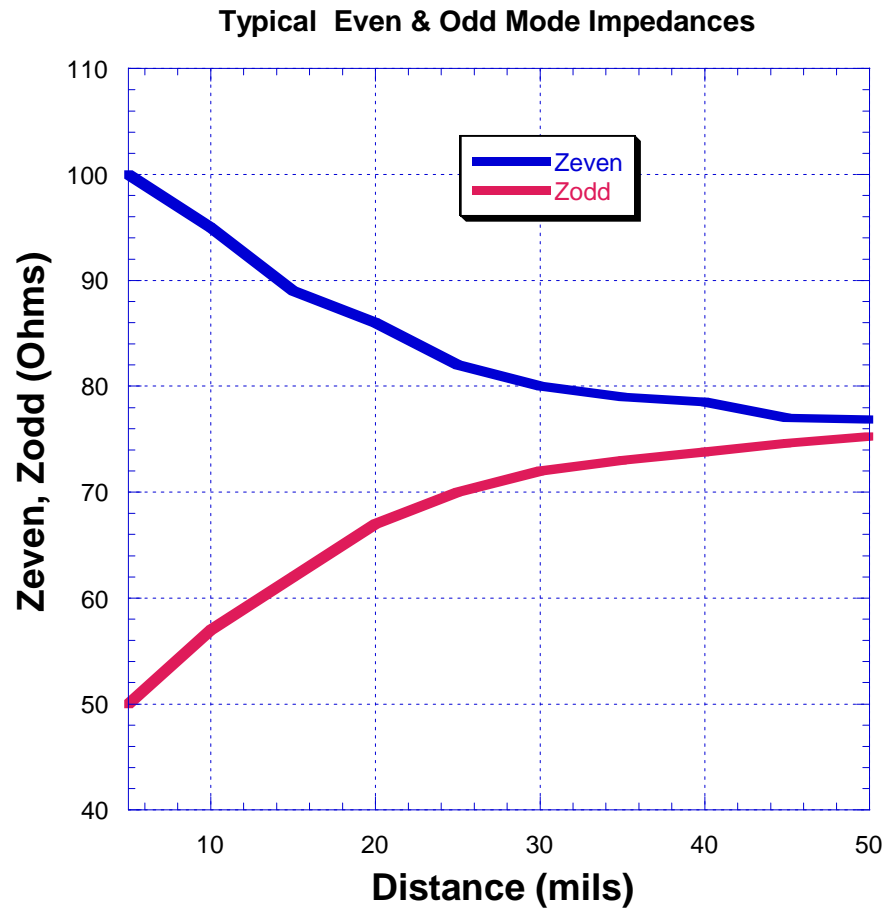
Measured mutual inductance



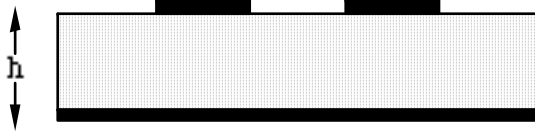
Measured mutual capacitance



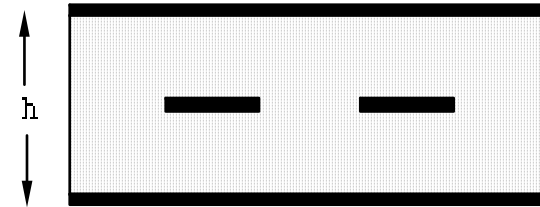
Even & Odd Mode Impedances



Modal Velocities in Stripline and Microstrip

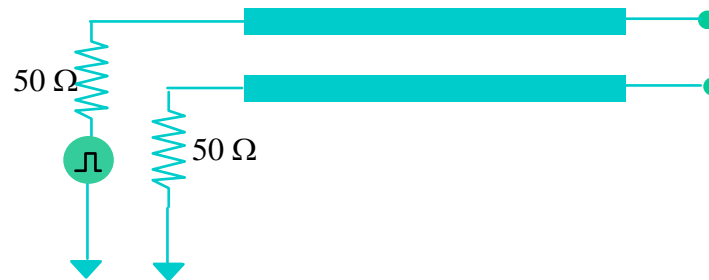


Microstrip : Inhomogeneous structure, odd and even-mode velocities must have different values.



Stripline : Homogeneous configuration, odd and even-mode velocities have approximately the same values.

Microstrip vs Stripline



Microstrip (h = 8 mils)

$$w = 8 \text{ mils}$$

$$\epsilon_r = 4.32$$

$$L_s = 377 \text{ nH/m}$$

$$C_s = 82 \text{ pF/m}$$

$$L_m = 131 \text{ nH/m}$$

$$C_m = 23 \text{ pF/m}$$

$$v_e = 0.155 \text{ m/ns}$$

$$v_d = 0.178 \text{ m/ns}$$

Stripline (h = 16 mils)

$$w = 8 \text{ mils}$$

$$\epsilon_r = 4.32$$

$$L_s = 466 \text{ nH/m}$$

$$C_s = 86 \text{ pF/m}$$

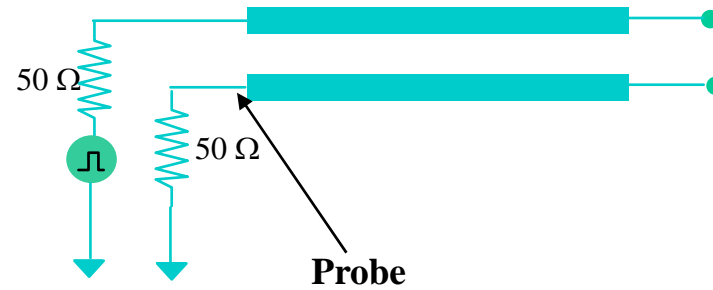
$$L_m = 109 \text{ nH/m}$$

$$C_m = 26 \text{ pF/m}$$

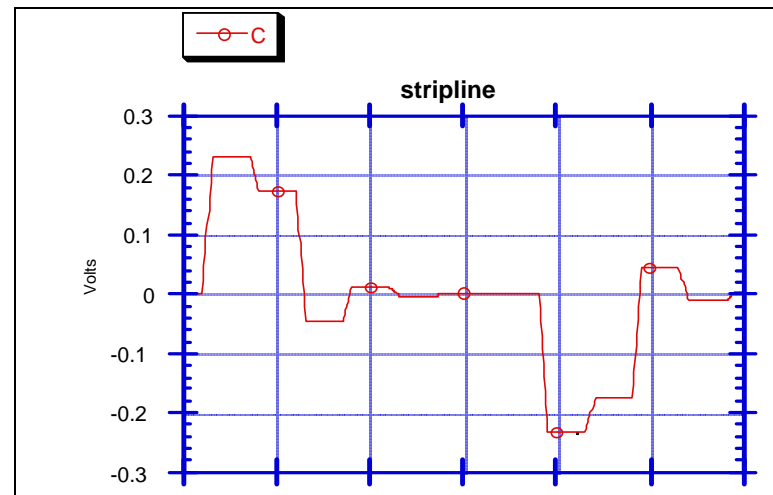
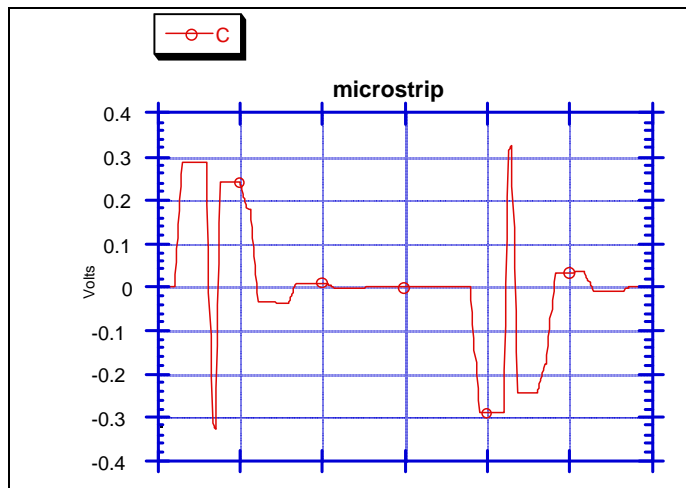
$$v_e = 0.142 \text{ m/ns}$$

$$v_d = 0.142 \text{ m/ns}$$

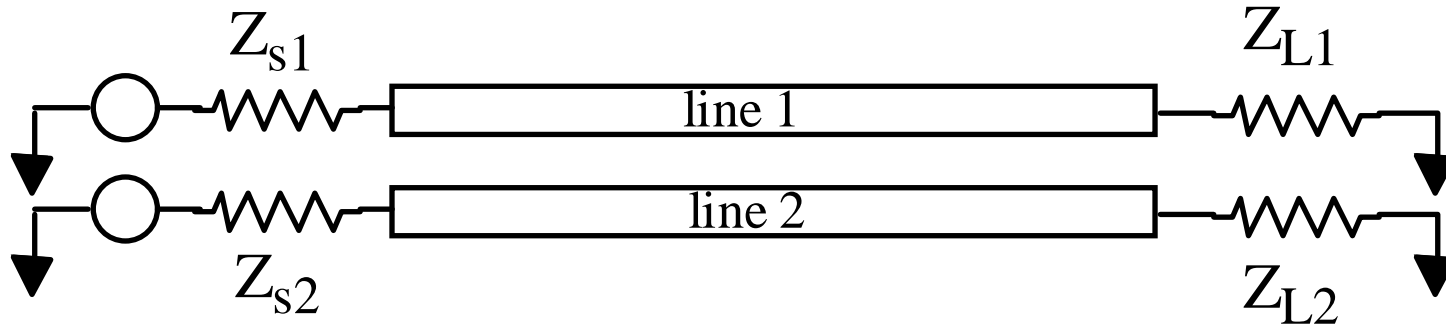
Microstrip vs Stripline



Sense line response at near end



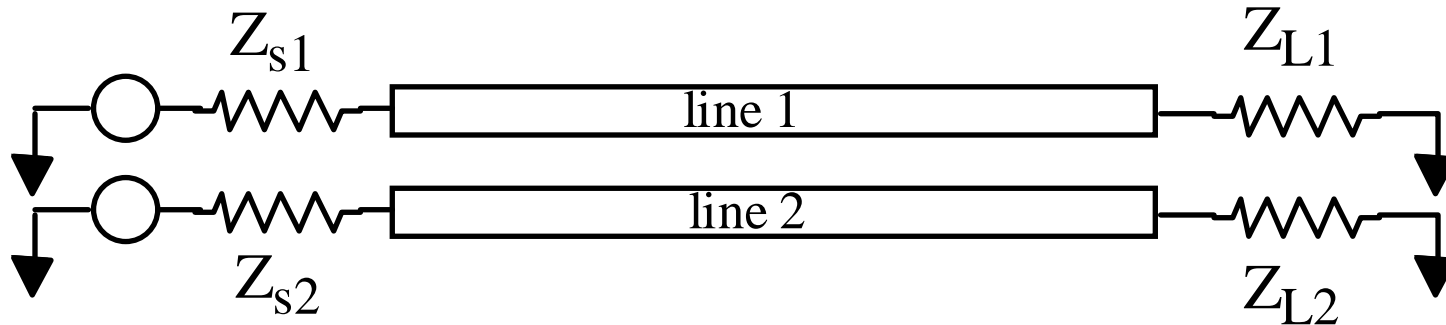
General Solution for Voltages



$$V_1(z) = \underbrace{A_e e^{-\frac{j\omega z}{v_e}} + B_e e^{+\frac{j\omega z}{v_e}}}_{\text{even}} + \underbrace{A_d e^{-\frac{j\omega z}{v_d}} + B_d e^{+\frac{j\omega z}{v_d}}}_{\text{odd}}$$

$$V_2(z) = \underbrace{A_e e^{-\frac{j\omega z}{v_e}} + B_e e^{+\frac{j\omega z}{v_e}}}_{\text{even}} - \underbrace{A_d e^{-\frac{j\omega z}{v_d}} - B_d e^{+\frac{j\omega z}{v_d}}}_{\text{odd}}$$

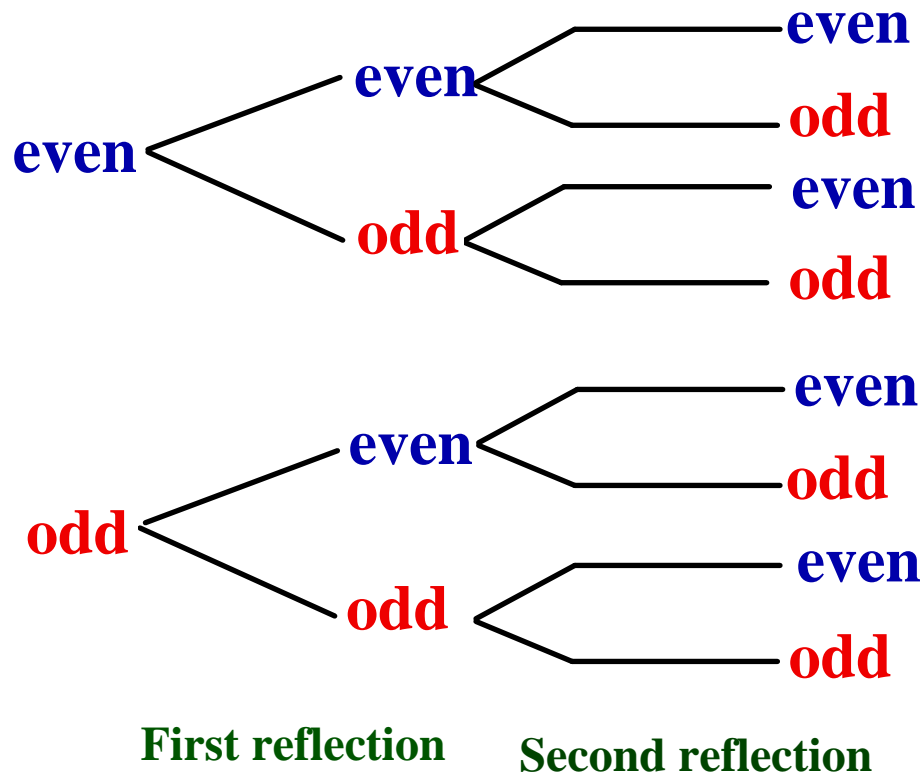
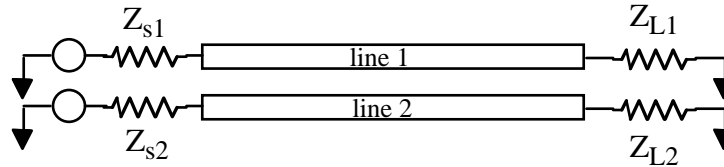
General Solution for Currents



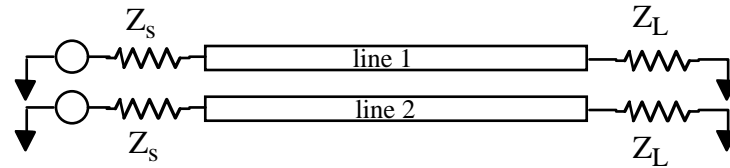
$$I_1(z) = \underbrace{\frac{1}{Z_e} \left[A_e e^{-\frac{j\omega z}{v_e}} - B_e e^{+\frac{j\omega z}{v_e}} \right]}_{\text{even}} + \underbrace{\frac{1}{Z_d} \left[A_d e^{-\frac{j\omega z}{v_d}} - B_d e^{+\frac{j\omega z}{v_d}} \right]}_{\text{odd}}$$

$$I_2(z) = \underbrace{\frac{1}{Z_e} \left[A_e e^{-\frac{j\omega z}{v_e}} - B_e e^{+\frac{j\omega z}{v_e}} \right]}_{\text{even}} - \underbrace{\frac{1}{Z_d} \left[A_d e^{-\frac{j\omega z}{v_d}} - B_d e^{+\frac{j\omega z}{v_d}} \right]}_{\text{odd}}$$

Coupling of Modes (asymmetric load)



Coupling of Modes (symmetric load)

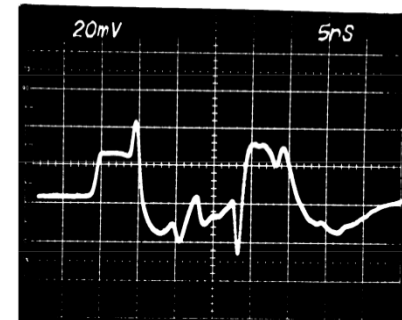
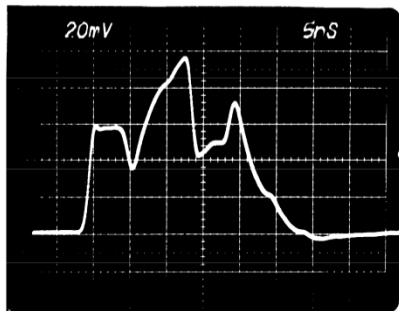
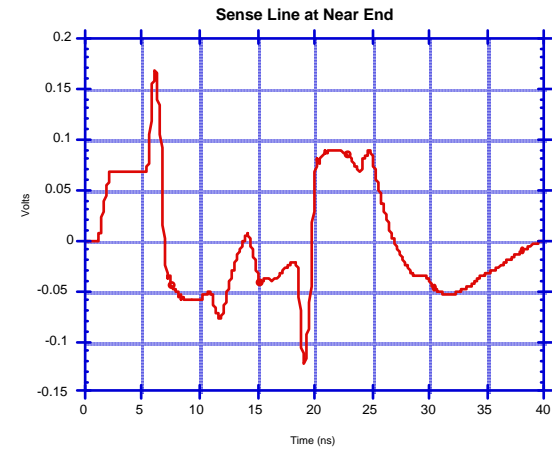
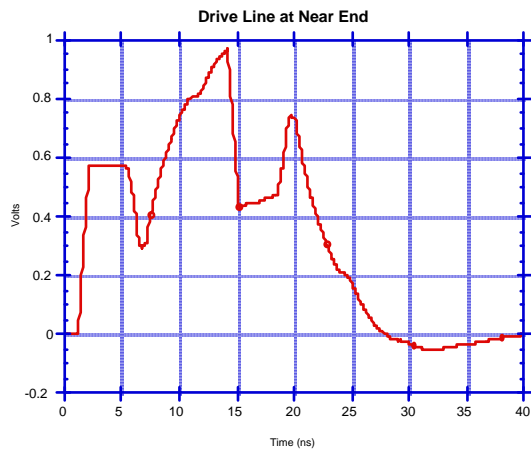
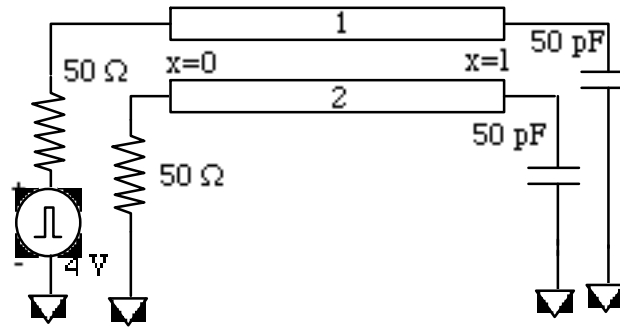


even ——— **even** ——— **even**

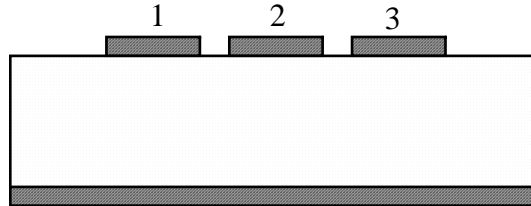
odd ——— **odd** ——— **odd**

First reflection

Second reflection



Three-Line Microstrip



$$-\frac{\partial V_1}{\partial z} = L_{11} \frac{\partial I_1}{\partial t} + L_{12} \frac{\partial I_2}{\partial t} + L_{13} \frac{\partial I_3}{\partial t}$$

$$-\frac{\partial I_1}{\partial z} = C_{11} \frac{\partial V_1}{\partial t} + C_{12} \frac{\partial V_2}{\partial t} + C_{13} \frac{\partial V_3}{\partial t}$$

$$-\frac{\partial V_2}{\partial z} = L_{21} \frac{\partial I_1}{\partial t} + L_{22} \frac{\partial I_2}{\partial t} + L_{23} \frac{\partial I_3}{\partial t}$$

$$-\frac{\partial I_2}{\partial z} = C_{21} \frac{\partial V_1}{\partial t} + C_{22} \frac{\partial V_2}{\partial t} + C_{23} \frac{\partial V_3}{\partial t}$$

$$-\frac{\partial V_3}{\partial z} = L_{31} \frac{\partial I_1}{\partial t} + L_{32} \frac{\partial I_2}{\partial t} + L_{33} \frac{\partial I_3}{\partial t}$$

$$-\frac{\partial I_3}{\partial z} = C_{31} \frac{\partial V_1}{\partial t} + C_{32} \frac{\partial V_2}{\partial t} + C_{33} \frac{\partial V_3}{\partial t}$$

Three-Line – Alpha Mode

Subtract (1c) from (1a) and (2c) from (2a), we get

$$\begin{aligned} -\frac{\partial V_\alpha}{\partial z} &= (L_{11} - L_{13}) \frac{\partial I_\alpha}{\partial t} \\ -\frac{\partial I_\alpha}{\partial z} &= (C_{11} - C_{13}) \frac{\partial V_\alpha}{\partial t} \end{aligned}$$

This defines the Alpha mode with:

$$V_\alpha = V_1 - V_3 \quad \text{and} \quad I_\alpha = I_1 - I_3$$

The wave impedance of that mode is:

$$Z_\alpha = \sqrt{\frac{L_{11} - L_{13}}{C_{11} - C_{13}}}$$

and the velocity is

$$u_\alpha = \frac{1}{\sqrt{(L_{11} - L_{13})(C_{11} - C_{13})}}$$

Three-Line – Modal Decomposition

In order to determine the next mode, assume that

$$V_\beta = V_1 + \beta V_2 + V_3$$

$$I_\beta = I_1 + \beta I_2 + I_3$$

$$-\frac{\partial V_\beta}{\partial z} = (L_{11} + \beta L_{21} + L_{31}) \frac{\partial I_1}{\partial t} + (L_{12} + \beta L_{22} + L_{32}) \frac{\partial I_2}{\partial t} + (L_{13} + \beta L_{23} + L_{33}) \frac{\partial I_3}{\partial t}$$
$$-\frac{\partial I_\beta}{\partial z} = (C_{11} + \beta C_{21} + C_{31}) \frac{\partial V_1}{\partial t} + (C_{12} + \beta C_{22} + C_{32}) \frac{\partial V_2}{\partial t} + (C_{13} + \beta C_{23} + C_{33}) \frac{\partial V_3}{\partial t}$$

By reciprocity $L_{32} = L_{23}$, $L_{21} = L_{12}$, $L_{13} = L_{31}$

By symmetry, $L_{12} = L_{23}$

Also by approximation, $L_{22} \approx L_{11}$, $L_{11} + L_{13} \approx L_{11}$

Three-Line – Modal Decomposition

$$-\frac{\partial V_\beta}{\partial z} = (L_{11} + \beta L_{12} + L_{13}) \left(\frac{\partial I_1}{\partial t} + \frac{\partial I_3}{\partial t} \right) + (2L_{12} + \beta L_{11}) \frac{\partial I_2}{\partial t}$$

In order to balance the right-hand side into I_β , we need to have

$$(2L_{12} + \beta L_{11}) I_2 = \beta (L_{11} + \beta L_{12} + L_{13}) I_2 \approx \beta (L_{11} + \beta L_{12}) I_2$$

$$2L_{12} = \beta^2 L_{12}$$

or $\beta = \pm\sqrt{2}$

Therefore the other two modes are defined as

The Beta mode with

Three-Line – Beta Mode

The Beta mode with

$$V_{\beta} = V_1 + \sqrt{2}V_2 + V_3$$

$$I_{\beta} = I_1 + \sqrt{2}I_2 + I_3$$

The characteristic impedance of the Beta mode is:

$$Z_{\beta} = \sqrt{\frac{L_{11} + \sqrt{2}L_{12} + L_{13}}{C_{11} + \sqrt{2}C_{12} + C_{13}}}$$

and propagation velocity of the Beta mode is

$$u_{\beta} = \frac{1}{\sqrt{(L_{11} + \sqrt{2}L_{12} + L_{13})(C_{11} + \sqrt{2}C_{12} + C_{13})}}$$

Three-Line – Delta Mode

The Delta mode is defined such that

$$V_{\delta} = V_1 - \sqrt{2}V_2 + V_3$$

$$I_{\delta} = I_1 - \sqrt{2}I_2 + I_3$$

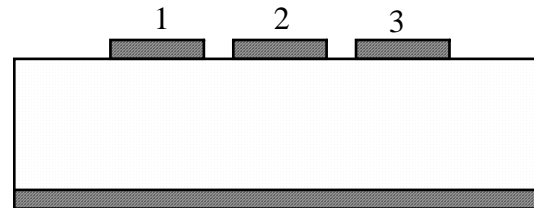
The characteristic impedance of the Delta mode is

$$Z_{\delta} = \sqrt{\frac{L_{11} - \sqrt{2}L_{12} + L_{13}}{C_{11} - \sqrt{2}C_{12} + C_{13}}}$$

The propagation velocity of the Delta mode is:

$$u_{\delta} = \frac{1}{\sqrt{(L_{11} - \sqrt{2}L_{12} + L_{13})(C_{11} - \sqrt{2}C_{12} + C_{13})}}$$

Symmetric 3-Line Microstrip



In summary: we have 3 modes for the 3-line system

$$E = \begin{pmatrix} 1 & 0 & -1 \\ 1 & \sqrt{2} & 1 \\ 1 & -\sqrt{2} & 1 \end{pmatrix}$$

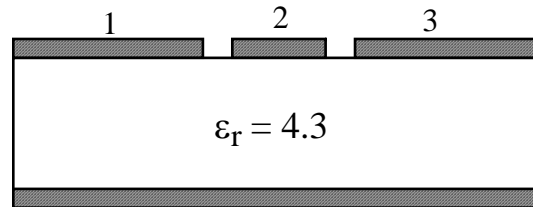
Alpha mode

Beta mode*

Delta mode*

*neglecting coupling between nonadjacent lines

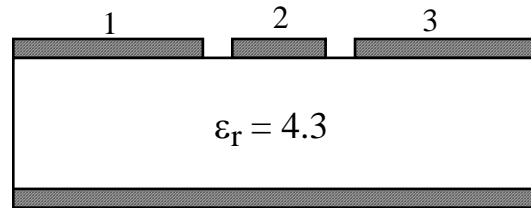
Coplanar Waveguide



$$L(nH / m) = \begin{pmatrix} 346 & 162 & 67 \\ 152 & 683 & 152 \\ 67 & 162 & 346 \end{pmatrix} \quad C(pF / m) = \begin{pmatrix} 113 & 17 & 5 \\ 16 & 53 & 16 \\ 5 & 17 & 113 \end{pmatrix}$$

$$E = \begin{pmatrix} 0.45 & 0.12 & 0.45 \\ 0.5 & 0 & -0.5 \\ -0.45 & 0.87 & -0.45 \end{pmatrix} \quad H = \begin{pmatrix} 0.44 & 0.49 & 0.44 \\ 0.5 & 0 & -0.5 \\ -0.10 & 0.88 & -0.10 \end{pmatrix}$$

Coplanar Waveguide

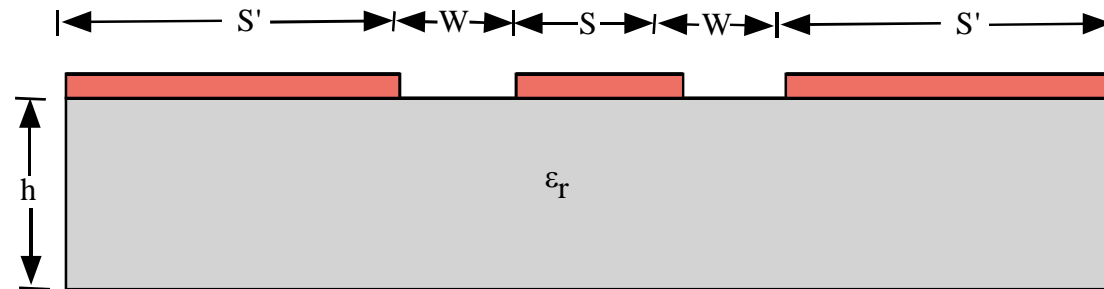


$$Z_m(\Omega) = \begin{pmatrix} 73 & 0 & 0 \\ 0 & 48 & 0 \\ 0 & 0 & 94 \end{pmatrix}$$

$$Z_c(\Omega) = \begin{pmatrix} 56 & 23 & 8 \\ 22 & 119 & 22 \\ 8 & 23 & 56 \end{pmatrix}$$

$$v_p(m/ns) = \begin{pmatrix} 0.15 & 0 & 0 \\ 0 & 0.17 & 0 \\ 0 & 0 & 0.18 \end{pmatrix}$$

Coplanar Waveguide



$K(k)$: Complete Elliptic Integral of the first kind

$$k = \frac{S}{S + 2W}$$

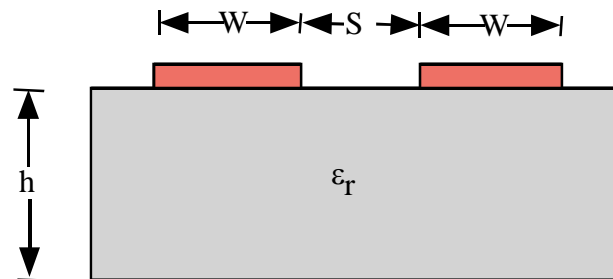
$$Z_{ocp} = \frac{30\pi}{\sqrt{\frac{\epsilon_r + 1}{2}}} \frac{K'(k)}{K(k)} \text{ (ohm)}$$

$$K'(k) = K(k')$$

$$k' = (1 - k^2)^{1/2}$$

$$v_{cp} = \left(\frac{2}{\epsilon_r + 1} \right)^{1/2} c$$

Coplanar Strips



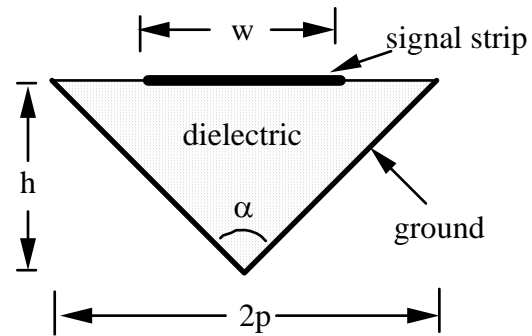
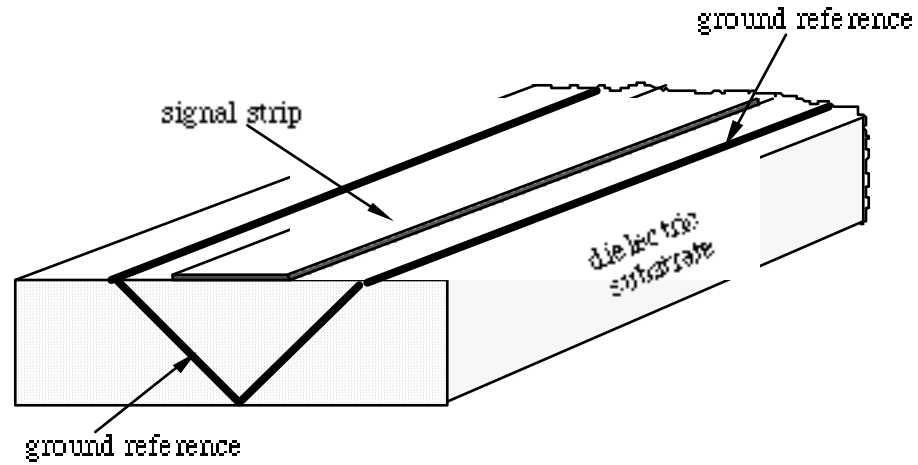
$$Z_{ocs} = \frac{120\pi}{\sqrt{\frac{\epsilon_r + 1}{2}}} \frac{K'(k)}{K(k)} \text{ (ohm)}$$

Qualitative Comparison

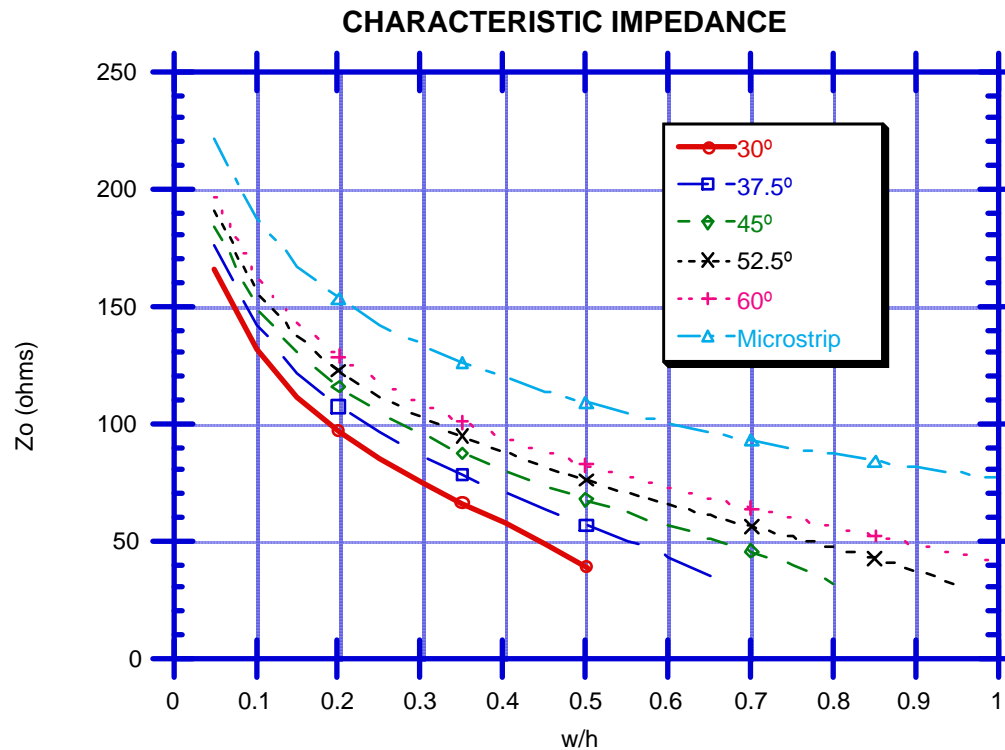
Characteristic	Microstrip	Coplanar Wguide	Coplanar strips
ϵ_{eff}^*	~6.5	~5	~5
Power handling	High	Medium	Medium
Radiation loss	Low	Medium	Medium
Unloaded Q	High	Medium	Low or High
Dispersion	Small	Medium	Medium
Mounting (shunt)	Hard	Easy	Easy
Mounting (series)	Easy	Easy	Easy

* Assuming $\epsilon_r=10$ and $h=0.025$ inch

V Transmission Line

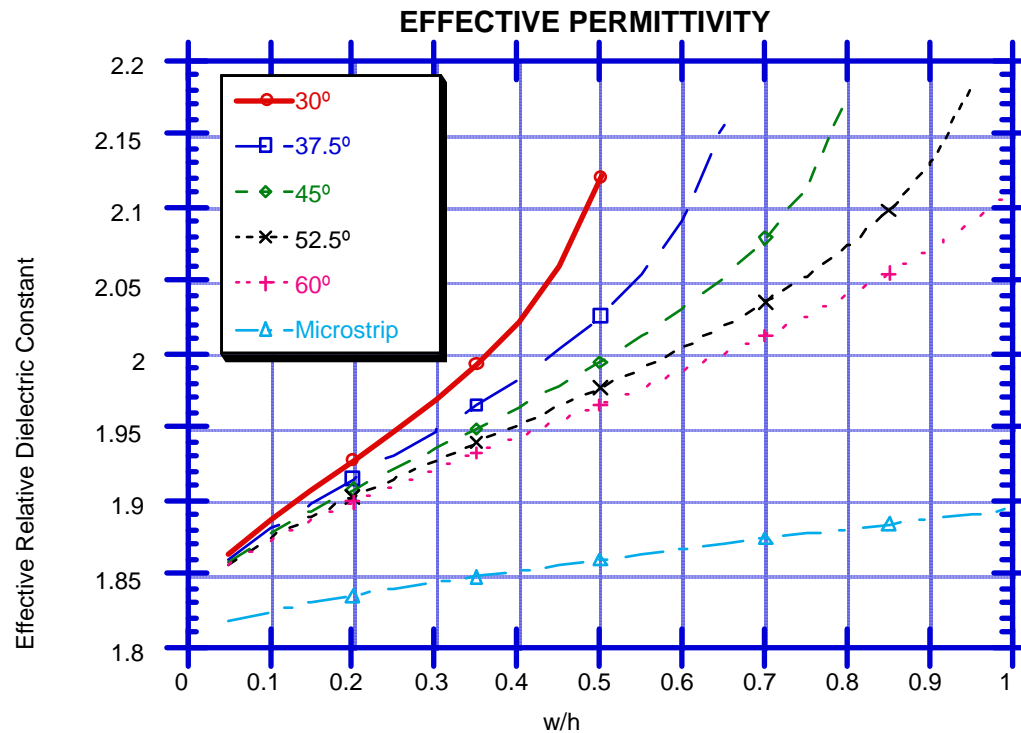


V-Line Characteristic Impedance



Calculated values of the characteristic impedance for a single-line v-strip structure as a function of width-to-height ratio w/h . The relative dielectric constant is $\epsilon_r = 2.55$.

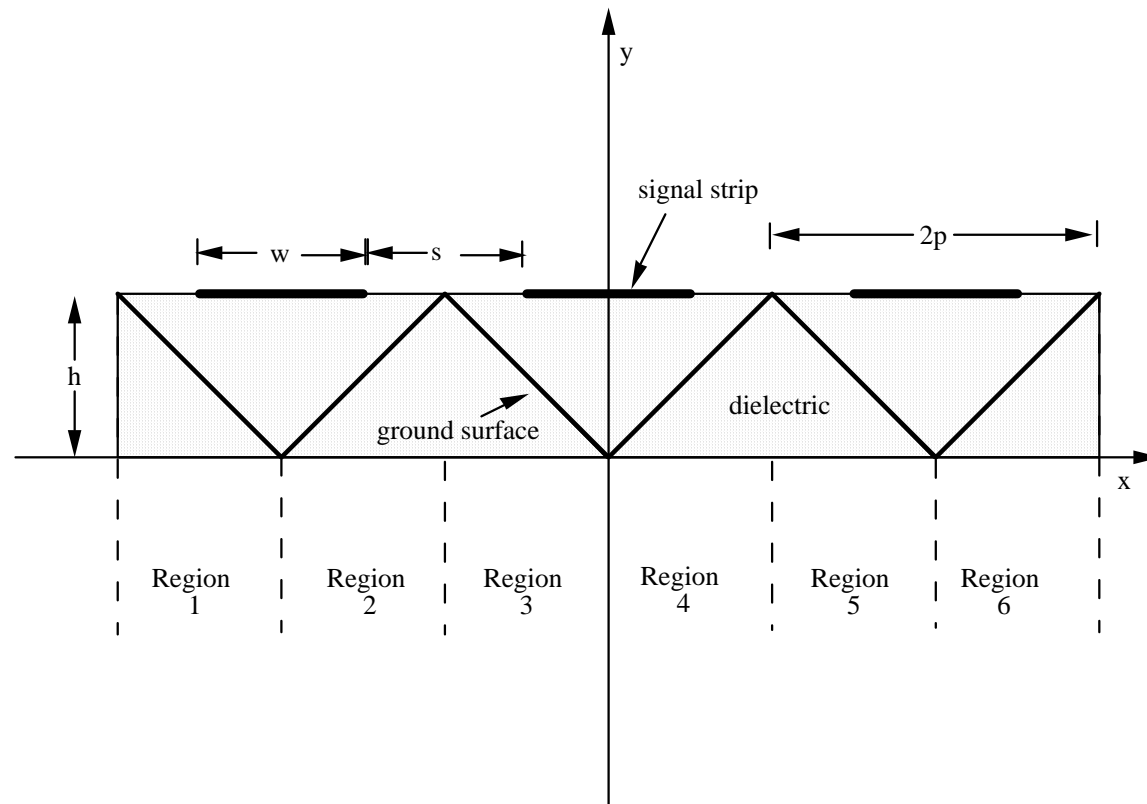
V-Line – Effective Permittivity



Calculated values of the effective relative dielectric constant for a single-line v-strip structure as a function of width-to-height ratio w/h .

The relative dielectric constant is $\epsilon_r = 2.55$

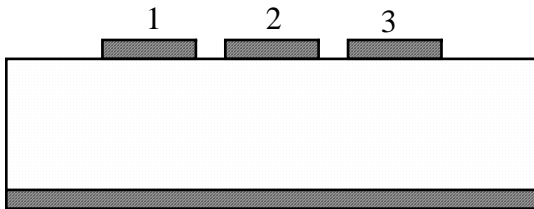
Three-Line – V Transmission Line



V-Line and Microstrip

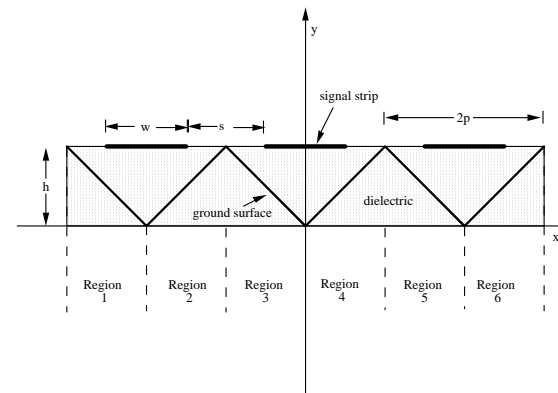
Microstrip

	52.49	-5.90	-0.57
C (pF/m) =	-5.90	53.27	-5.90
	-0.57	-5.90	52.49
	609.74	113.46	41.79
L (nH/m) =	113.54	607.67	113.54
	41.79	113.46	609.74



V-line

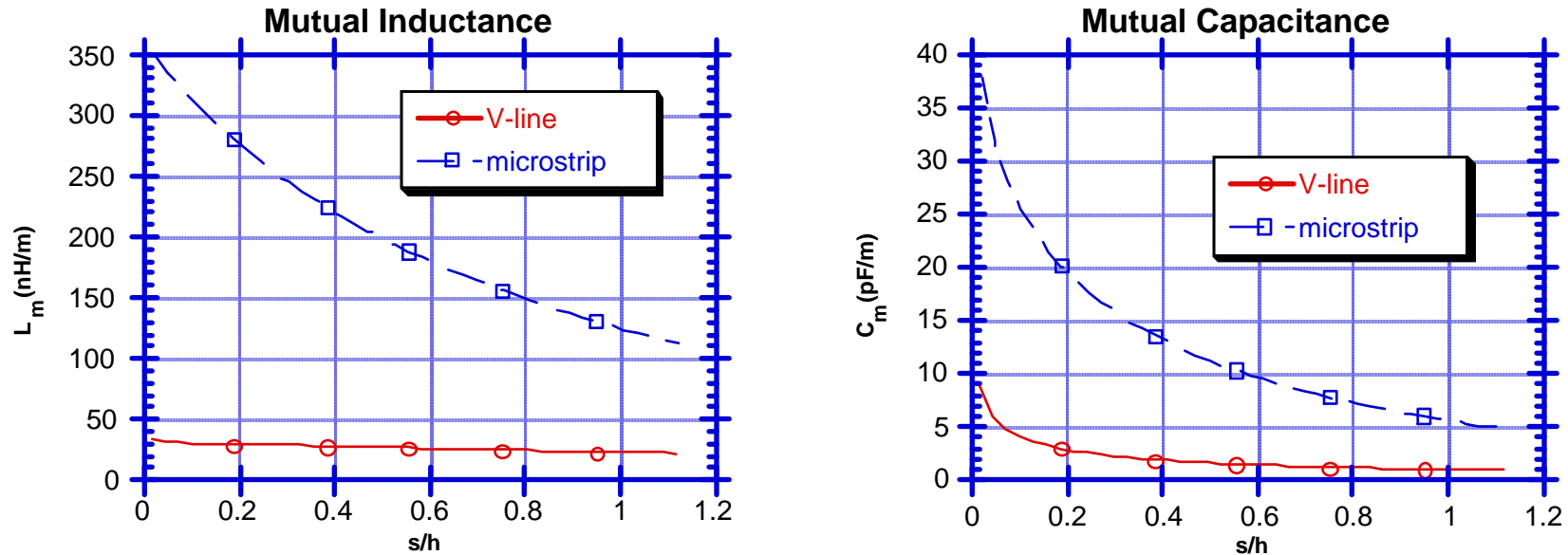
	74.00	-0.97	-0.23
C (pF/m) =	-0.97	74.07	-0.97
	-0.23	-0.97	74.00
	425.84	20.35	7.00
L (nH/m) =	20.36	422.85	20.36
	7.00	20.35	425.84



Comparison of the inductance and capacitance matrices between a three-line v-line and microstrip structures.

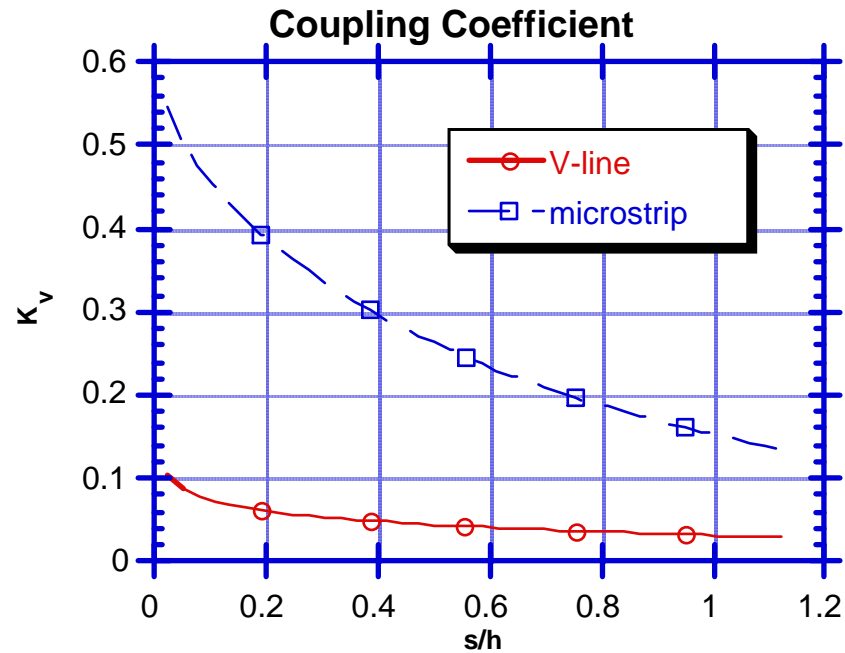
The parameters are $p/h = 0.8$ mils, $w/h = 0.6$ and $\epsilon_r = 4.0$.

V-Line: Coupling Coefficients



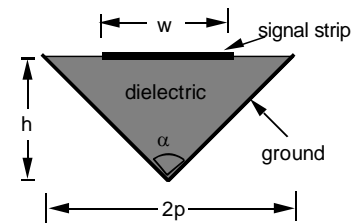
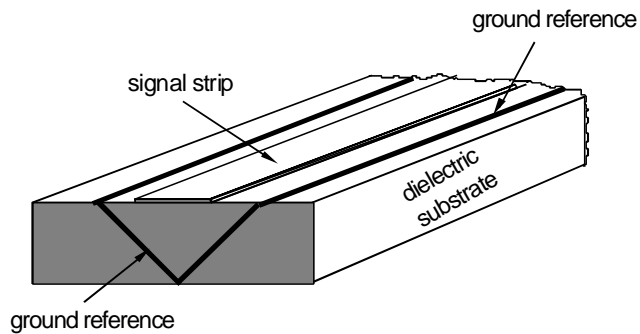
Plot of mutual inductance (top) and mutual capacitance (bottom) versus spacing-to-height ratio for v-line and microstrip configurations. The parameters are $w/h = 0.24$, $\epsilon_r = 4.0$.

V-Line vs Microstrip: Coupling Coefficients



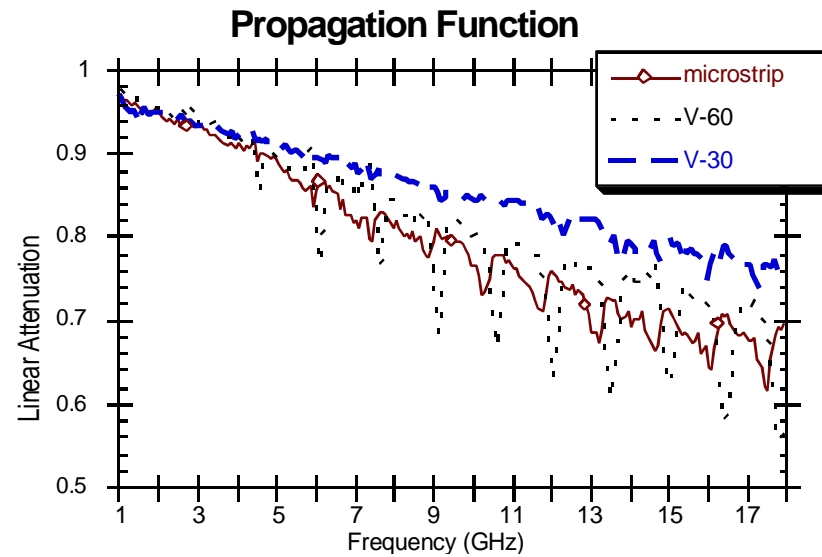
Plot of the coupling coefficient versus spacing-to-height ratio for v-line and microstrip configurations. The parameters are $w/h = 0.24$, $\epsilon_r = 4.0$.

Advantages of V-Line



Advantages of V Line

- * Higher bandwidth
- * Lower crosstalk
- * Better transition



V-Line vs Microstrip: Insertion Loss

