

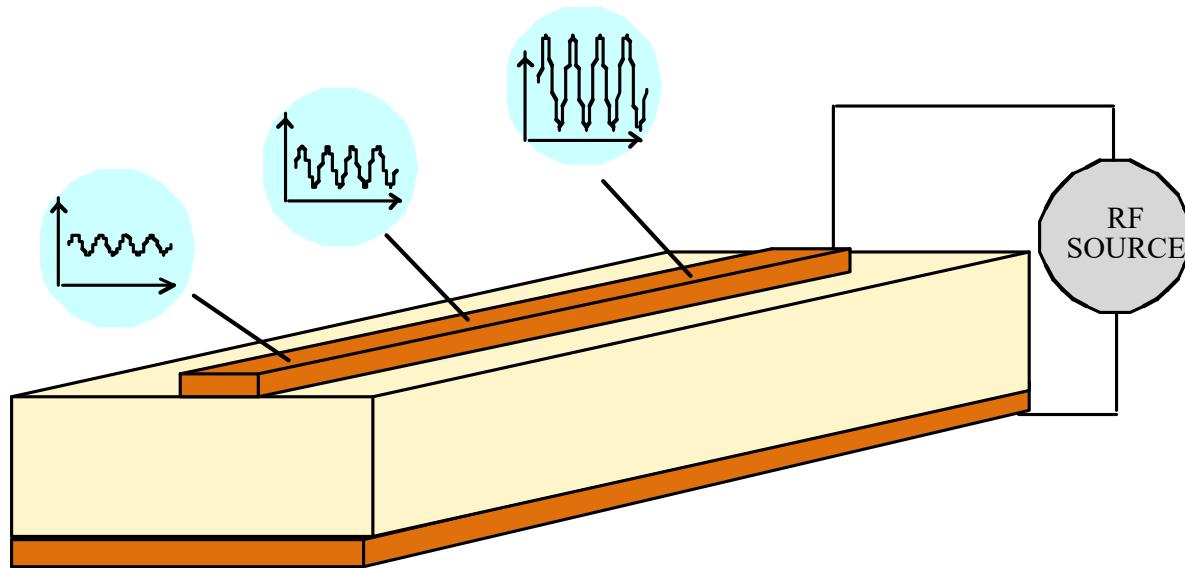
ECE 451

Advanced Microwave Measurements

Lossy Transmission Lines

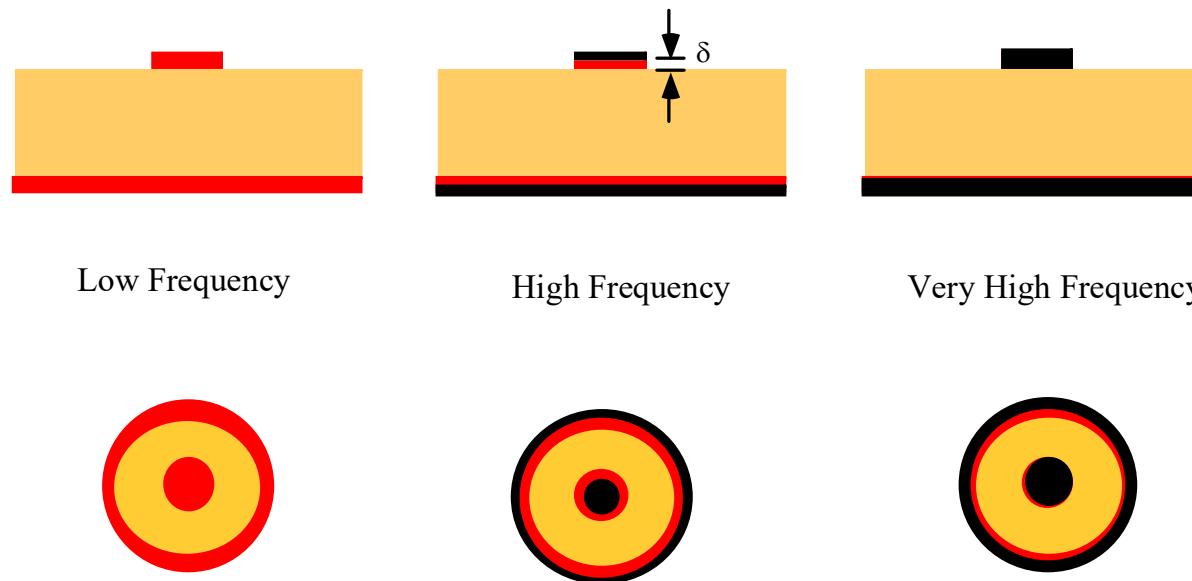
Jose E. Schutt-Aine
Electrical & Computer Engineering
University of Illinois
jschutt@emlab.uiuc.edu

Loss in Transmission Lines

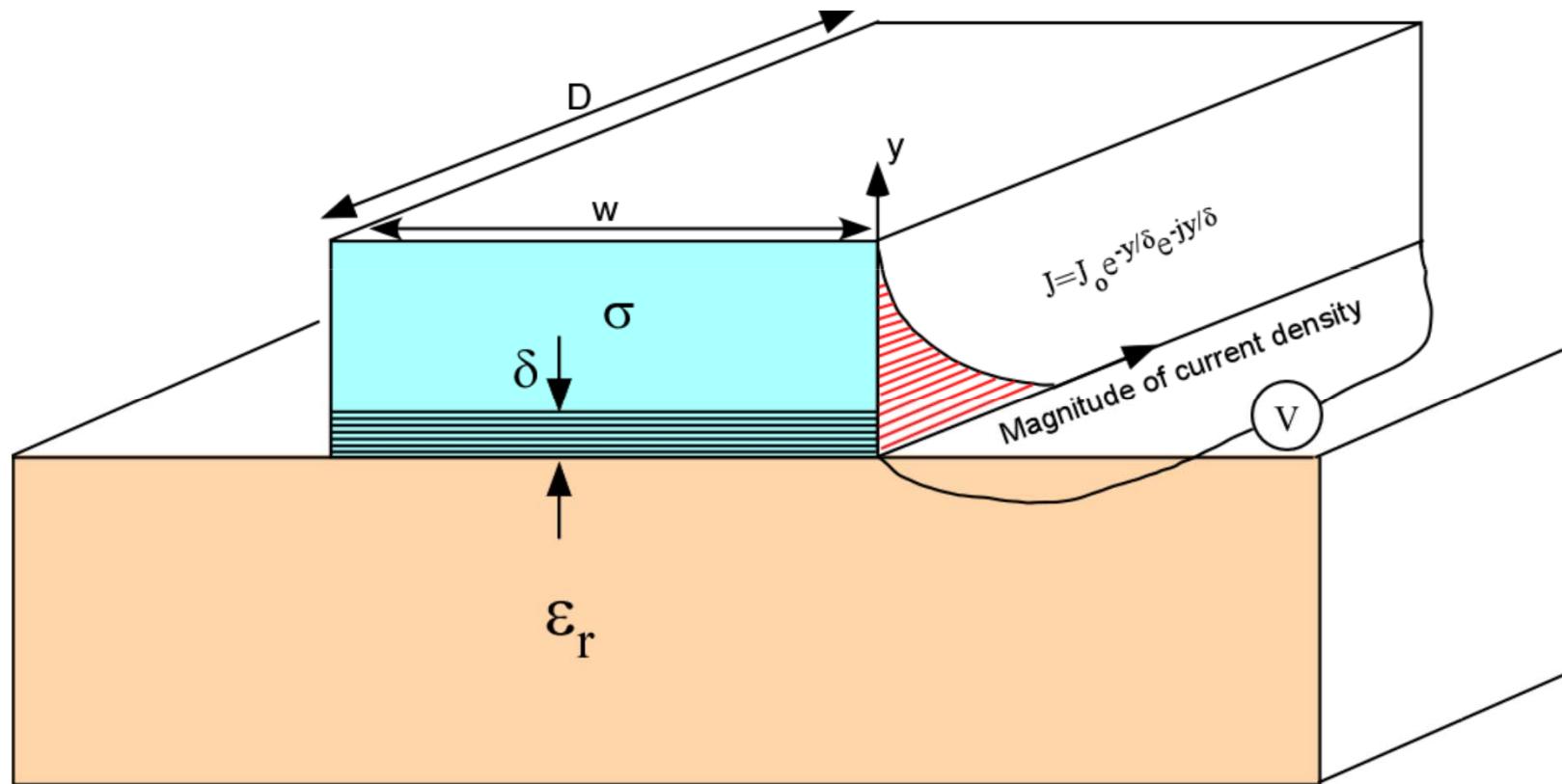


Signal amplitude decreases with distance from the source.

Skin Effect in Lines



Skin Effect in Microstrip



H. A. Wheeler, "Formulas for the skin effect," Proc. IRE, vol. 30, pp. 412-424, 1942

Skin Effect in Microstrip

Current density varies as

$$J = J_o e^{-y/\delta} e^{-jy/\delta}$$

Note that the phase of the current density varies as a function of y

$$I = \int_0^{\infty} J_o w e^{-y/\delta} e^{-jy/\delta} dy = \frac{J_o w \delta}{1 + j}$$

$$\sigma E_o = J_o \Rightarrow E_o = \frac{J_o}{\sigma}$$

The voltage measured over a section of conductor of length D is:

$$V = E_o D = \frac{J_o D}{\sigma}$$

Skin Effect in Microstrip

The skin effect impedance is

$$Z_{skin} = \frac{V}{I} = \frac{J_o D}{\sigma} \frac{(1+j)}{J_o w \delta} = \frac{D}{w} (1+j) \sqrt{\pi f \mu \rho}$$

where $\rho = \frac{1}{\sigma}$ is the bulk resistivity of the conductor

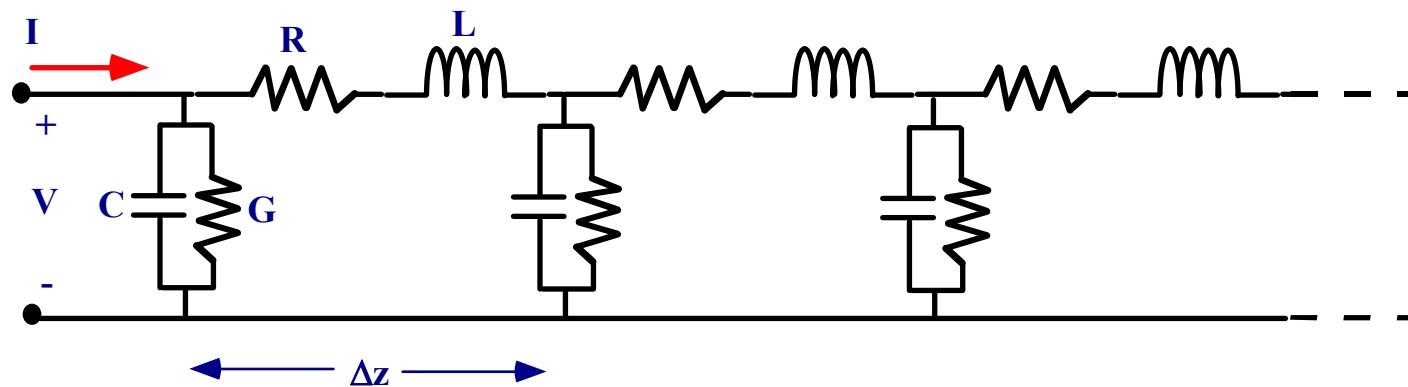
$$Z_{skin} = R_{skin} + jX_{skin}$$

with

$$R_{skin} = X_{skin} = \frac{D}{w} \sqrt{\pi f \mu \sigma}$$

→ Skin effect has reactive (inductive) component

Lossy Transmission Line

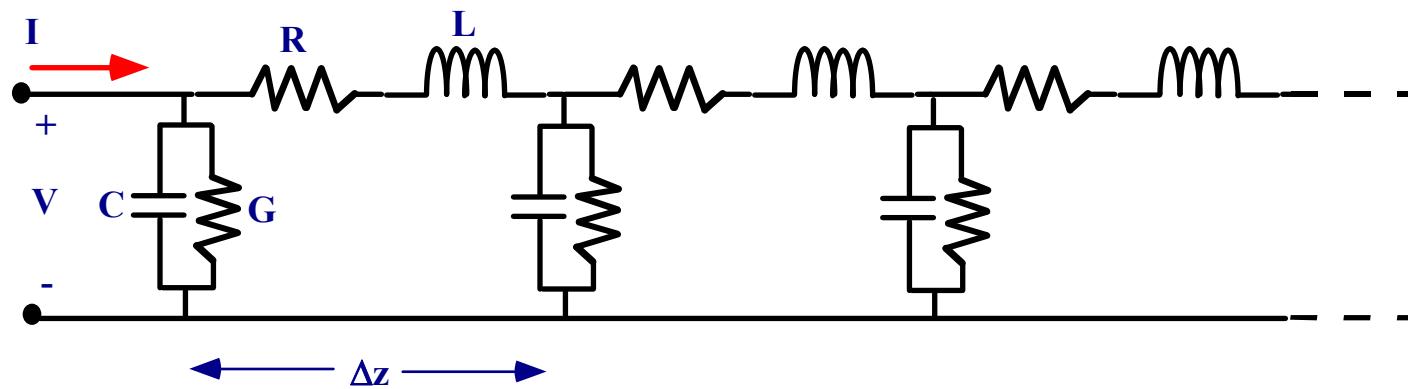


Telegraphers Equation: Time Domain

$$-\frac{\partial V}{\partial z} = RI + L \frac{\partial I}{\partial t}$$

$$-\frac{\partial I}{\partial z} = GV + C \frac{\partial V}{\partial t}$$

Lossy Transmission Line

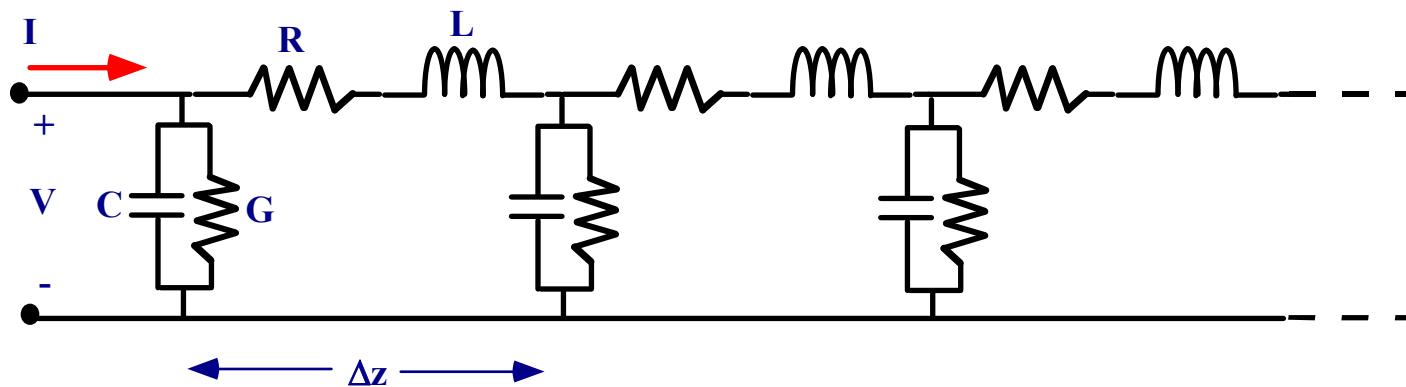


Telegraphers Equation: Frequency Domain

$$-\frac{\partial V}{\partial z} = (R + j\omega L)I = ZI$$

$$-\frac{\partial I}{\partial z} = (G + j\omega C)V = YV$$

Lossy Transmission Line

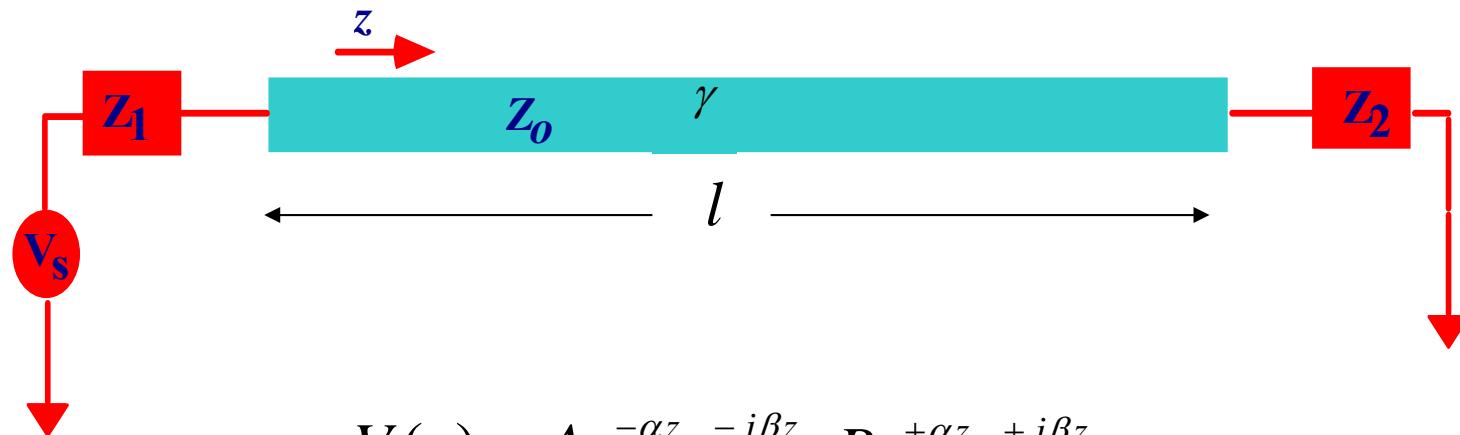


Telegraphers Equation: Frequency Domain

$$-\frac{\partial^2 V}{\partial z^2} = (R + j\omega L)(G + j\omega C)V = ZV = \gamma^2 V$$

$$-\frac{\partial^2 I}{\partial z^2} = (G + j\omega C)(R + j\omega L)I = YI = \gamma^2 I$$

Lossy Transmission Line



$$V(z) = Ae^{-\alpha z} e^{-j\beta z} + Be^{+\alpha z} e^{+j\beta z}$$

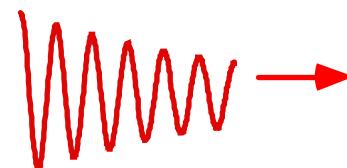
$$I(z) = \frac{1}{Z_o} \left[Ae^{-\alpha z} e^{-j\beta z} - Be^{+\alpha z} e^{+j\beta z} \right]$$

$$Z_o = \sqrt{\frac{(R + j\omega L)}{(G + j\omega C)}} \quad \gamma = \alpha + j\beta = \sqrt{(R + j\omega L)(G + j\omega C)}$$

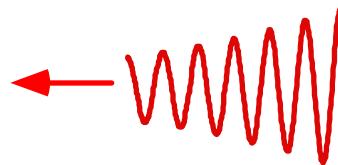
Lossy Transmission Line



forward wave

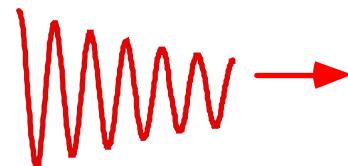


backward wave



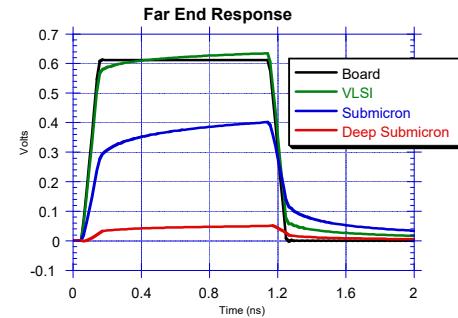
Effects of Losses

- Signal attenuation

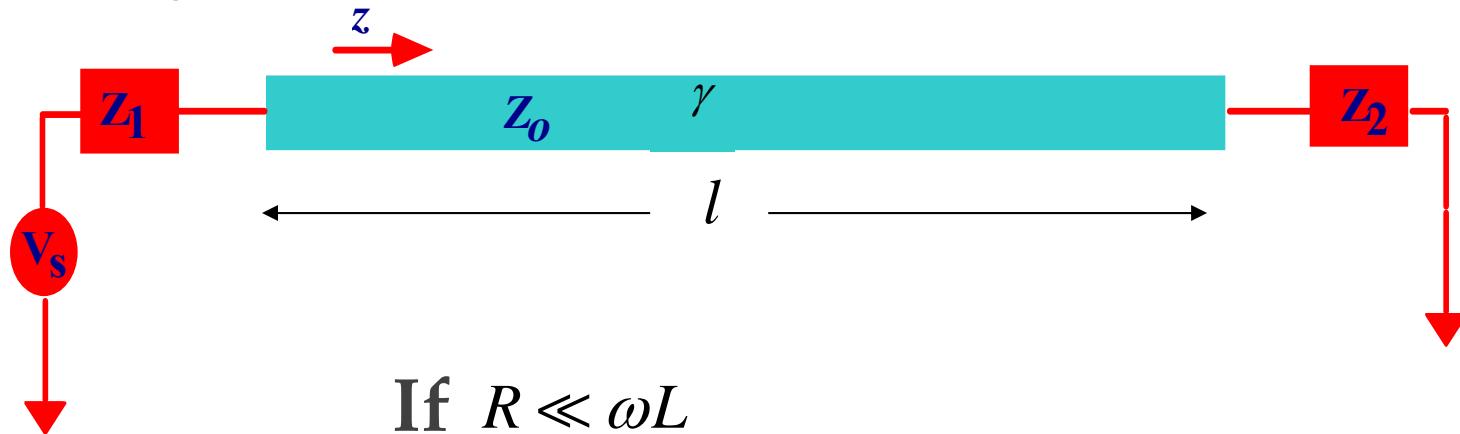


- Dispersion $\gamma = \alpha(\omega) + j\beta(\omega) = \sqrt{(R + j\omega L)(G + j\omega C)}$

- Rise time degradation



Special Case – Low Loss



If $R \ll \omega L$

and $G \ll \omega C$

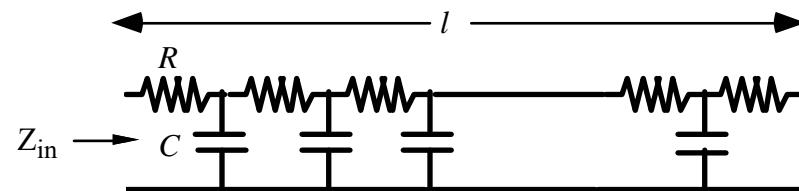
$$Z_o = \sqrt{\frac{j\omega L \left(1 + \frac{R}{j\omega L}\right)}{j\omega C \left(1 + \frac{G}{j\omega C}\right)}} \simeq \sqrt{\frac{L}{C}}$$

$$\gamma = \alpha + j\beta$$

$$\alpha \simeq \frac{1}{2} \left(R \sqrt{\frac{C}{L}} + G \sqrt{\frac{L}{C}} \right)$$

$$\beta \simeq \omega \sqrt{LC} \quad v_p = \frac{\omega}{\beta} \simeq \frac{1}{\sqrt{LC}}$$

RC Transmission Line



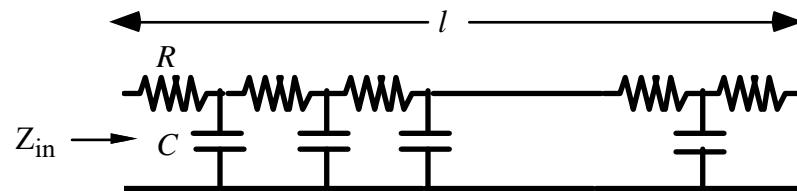
R : series resistance per unit length

C : shunt capacitance per unit length

$$Z_{in} = \frac{Rl \coth \frac{Rl}{\sqrt{2}} \sqrt{\frac{C\omega}{R}} (1+j)}{\frac{Rl}{\sqrt{2}} \sqrt{\frac{C\omega}{R}} (1+j)}$$

For very high ω , $\arg(Z_{in}) \approx 45^\circ$

RC Transmission Line



If $\omega \ll \frac{2}{RCl^2}$ then

$$Z_{in} = \frac{Rl}{2} + \frac{1}{jCl\omega} = \frac{R_T}{2} + \frac{1}{jC_T\omega}$$

$R_T = Rl$: total resistance

$C_T = Cl$: total capacitance

RC Transmission Line



Pulse Characteristics:

rise time: 100 ps

fall time: 100 ps

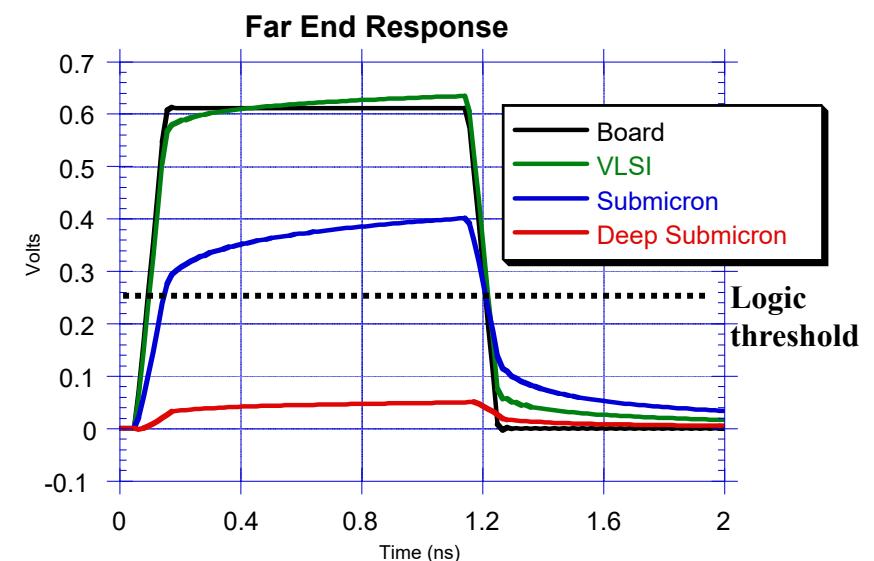
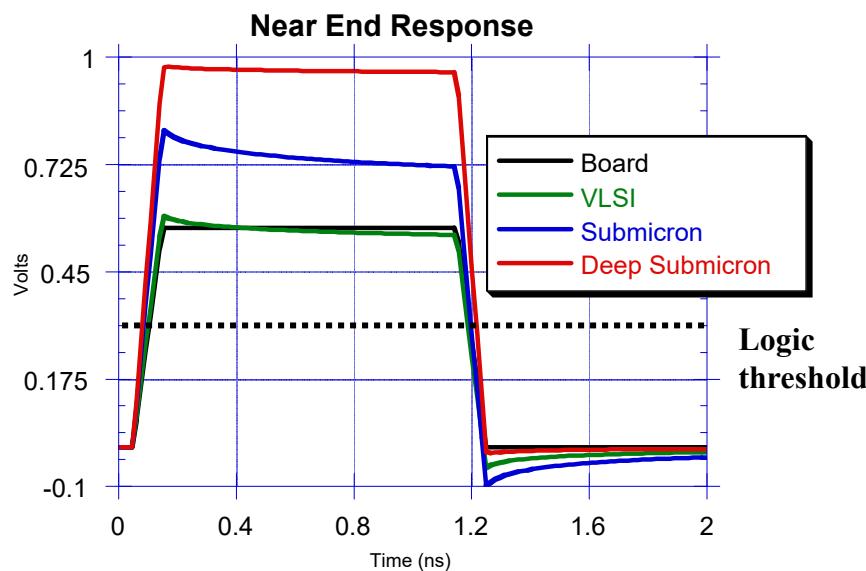
pulse width: 4ns

Line Characteristics

length : 3 mm

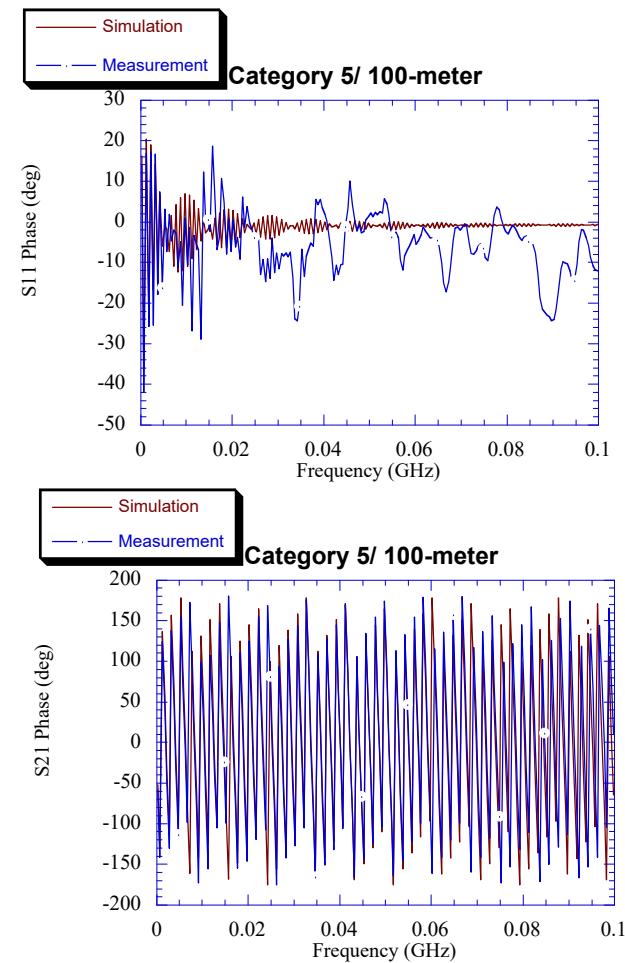
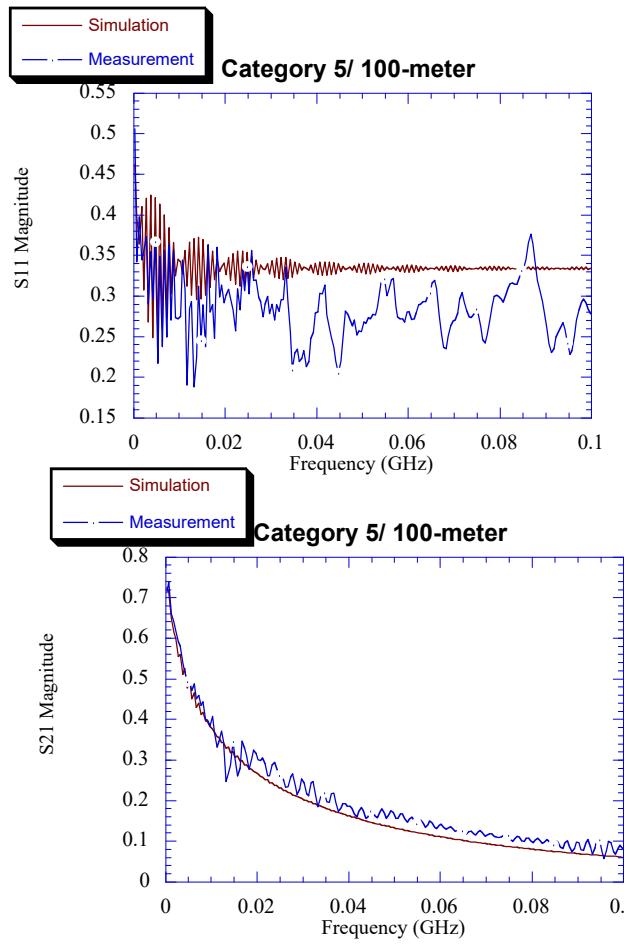
near end termination: 50Ω

far end termination 65Ω



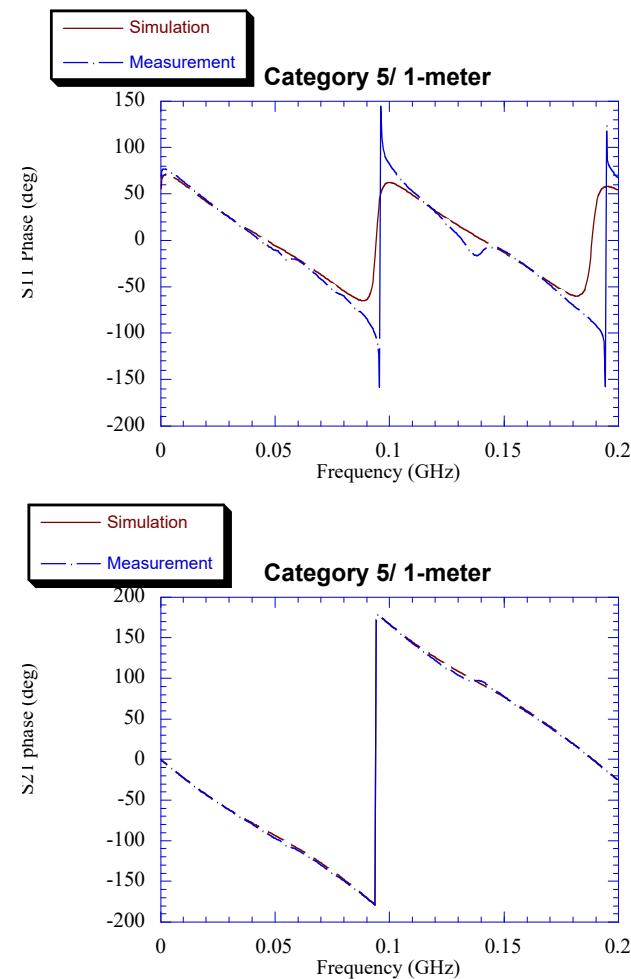
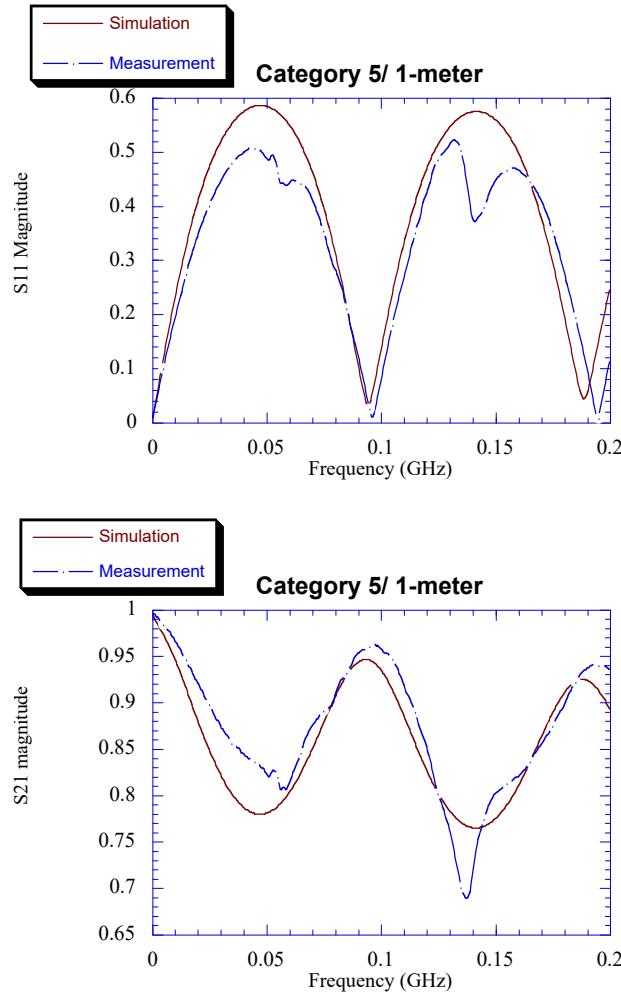
Long Cable

100m Category-5 Cable



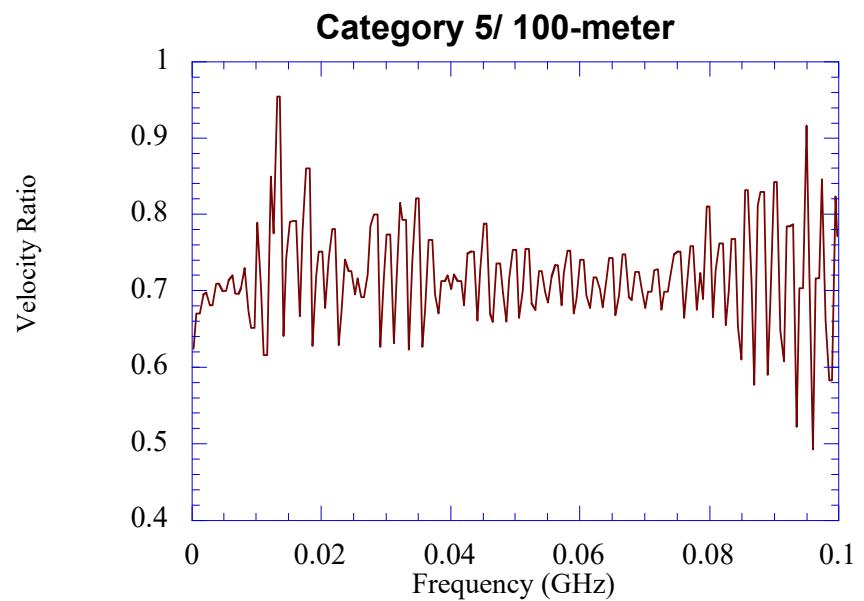
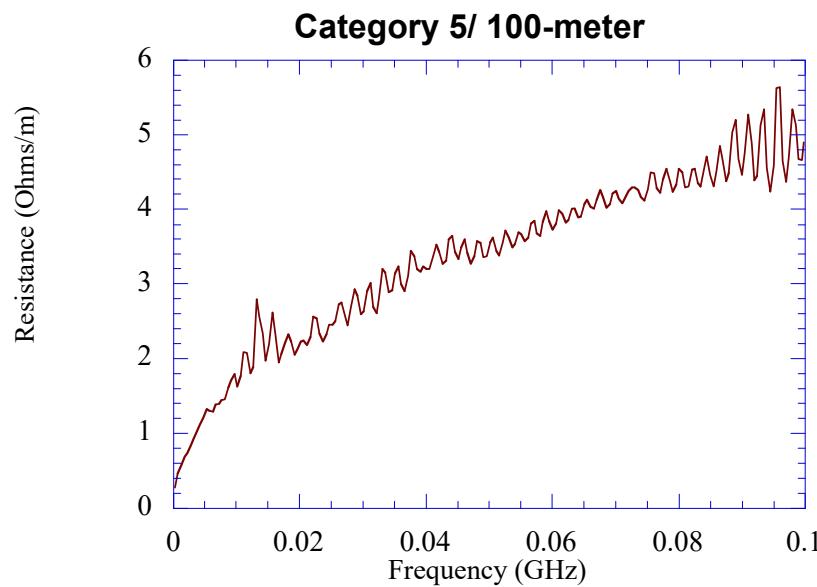
Short Cable

1m Category-5 Cable



Category 5 Cable

Resistance and velocity



Cable Loss Model

$$R(f) = R_s * f^p$$

$$\nu_r = \nu_{ro} + \nu_{rs} * f$$

$$Z = R(f) + j\omega L = R_{skin} + j(R_{skin} + \omega L)$$

	Z_0 (Ω)	V_{ro} (m/ns)	V_{rs} (m/ns-GHz)	R_s ($\Omega/m \cdot GHz^p$)	p	f_{max} (GHz)
Category 5	100	0.724	-0.165	15.38	0.482	0.2
24-Ga	100	0.678	1.157	29.03	0.593	0.1
Category 3	100	0.705	11.06	12.31	0.473	0.01
SMA	50	0.700	0.113	7.94	0.415	0.2

Lossy TL Simulation

- To simulate lossy TL with resistive loads
 - No closed form solution
 - Simplest method is to use IFFT

$$v(t, z) = \text{IFFT} \left\{ A e^{-\alpha z} e^{-j\beta z} + B e^{+\alpha z} e^{+j\beta z} \right\}$$

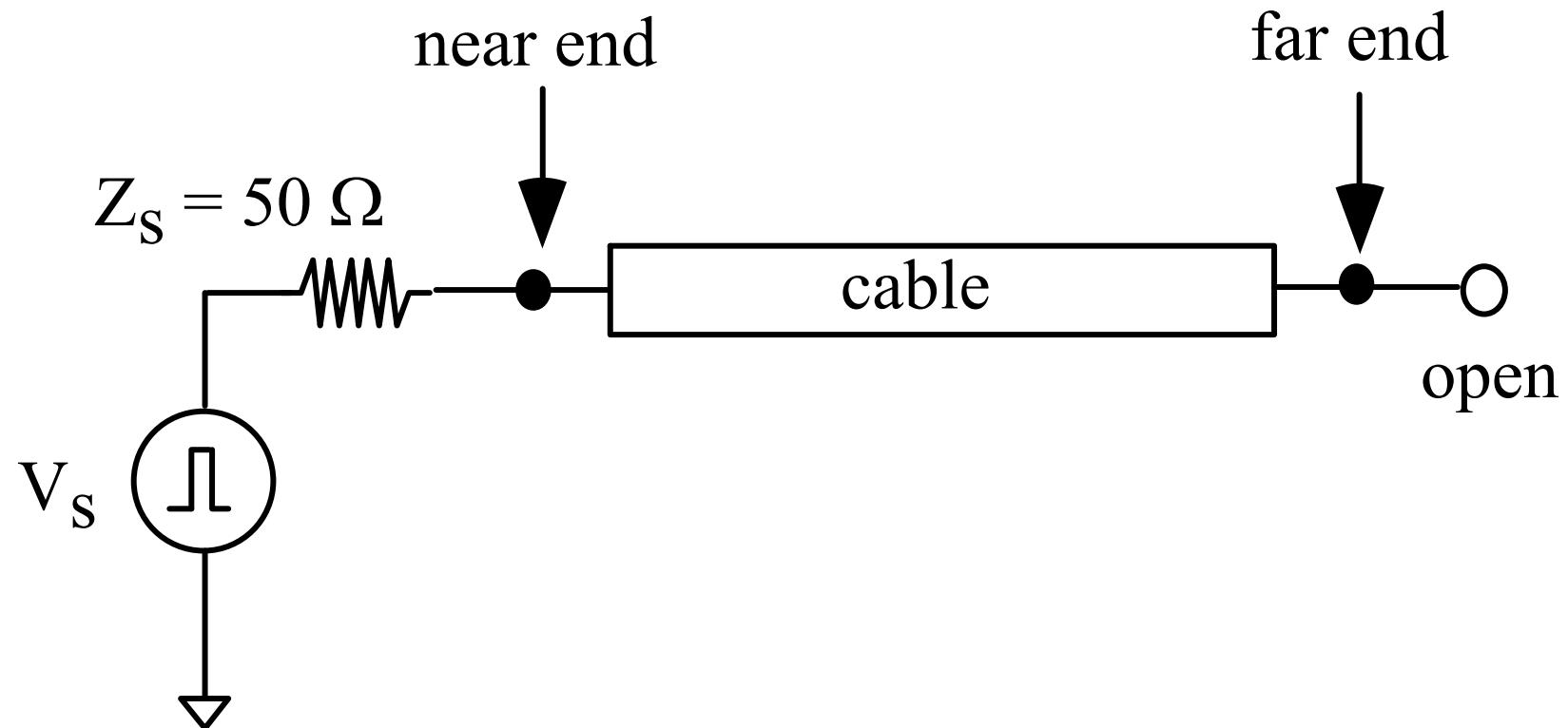
$$i(t, z) = \text{IFFT} \left\{ \frac{1}{Z_o} \left[A e^{-\alpha z} e^{-j\beta z} + B e^{+\alpha z} e^{+j\beta z} \right] \right\}$$

$$Z_o = \sqrt{\frac{(R + j\omega L)}{(G + j\omega C)}} \quad \gamma = \alpha + j\beta = \sqrt{(R + j\omega L)(G + j\omega C)}$$

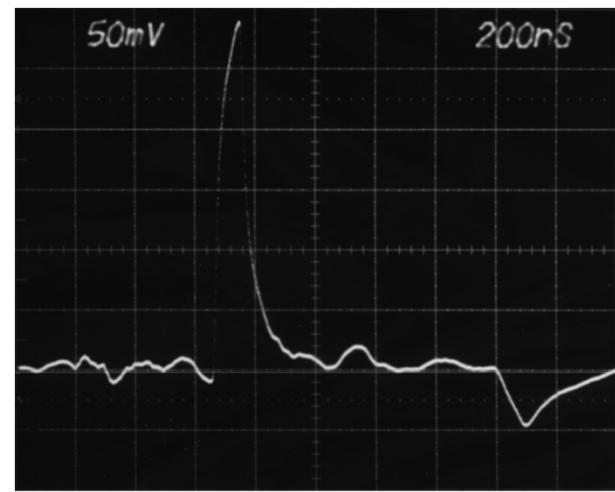
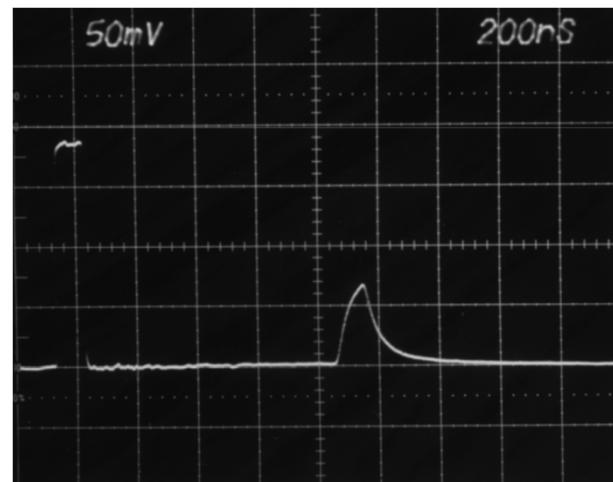
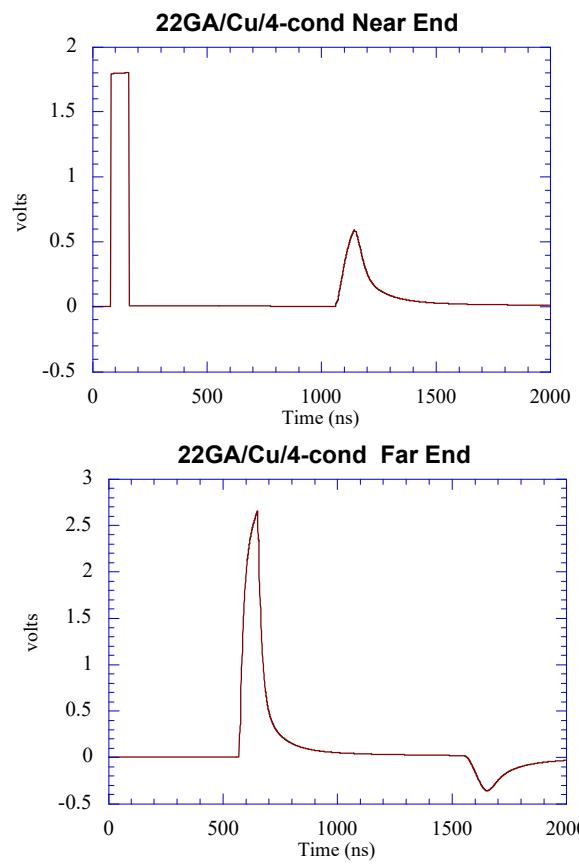
$$T = \frac{Z_o}{Z_1 + Z_o}$$

$$A = \frac{TV_s(\omega)}{1 - \Gamma_1 \Gamma_2 e^{-2\gamma l}} \quad B = \Gamma_2 e^{-2\gamma l} A \quad \Gamma_2 = \frac{Z_2 - Z_o}{Z_2 + Z_o} \quad \Gamma_1 = \frac{Z_1 - Z_o}{Z_1 + Z_o}$$

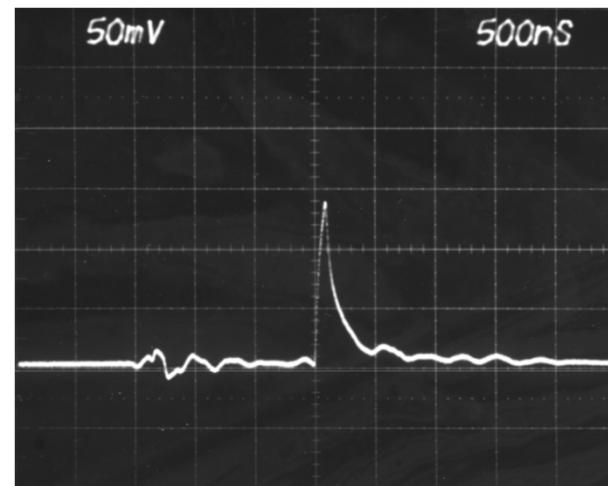
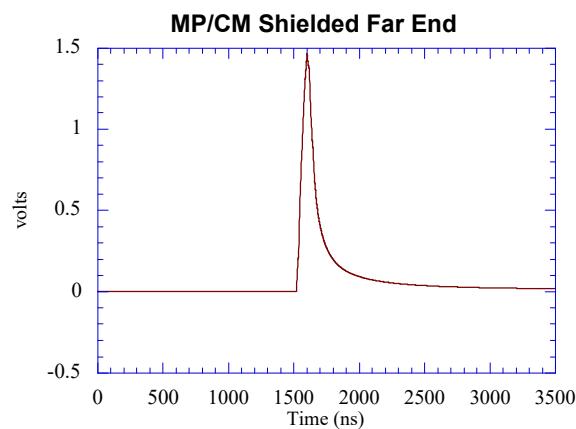
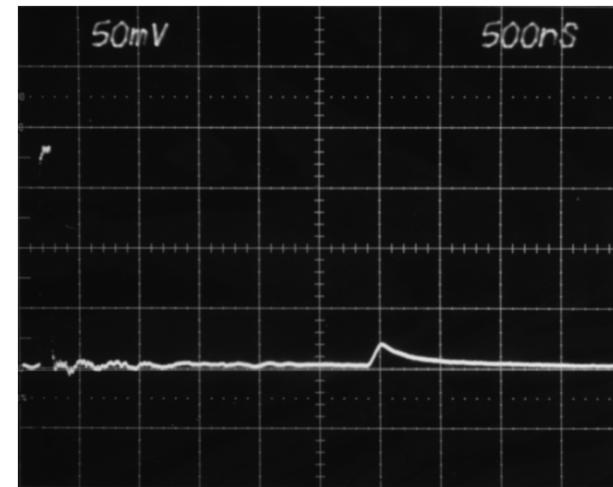
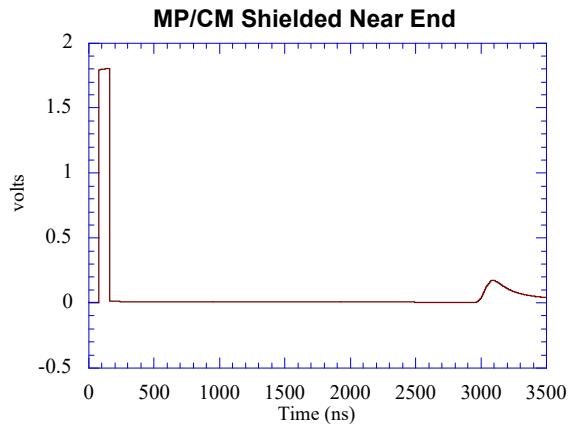
Time-Domain Simulations



Pulse Propagation (CAT-5)



Pulse Propagation (MP/CM)



Pulse Propagation (RG174)

