ECE 453 Wireless Communication Systems

Mixers

Jose E. Schutt-Aine
Electrical & Computer Engineering
University of Illinois
jesa@illinois.edu



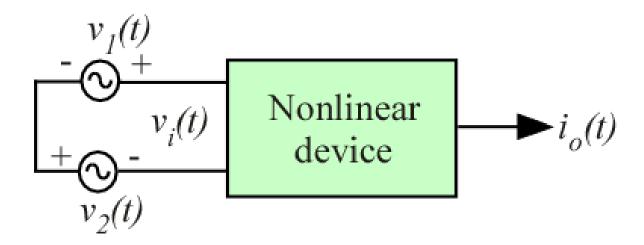
Mixers

Frequency translation modules used in receivers to produce baseband signals

- Frequency can be higher (up-conversion) or lower (down-conversion)
- Make use of nonlinearities in devices
- Can be based on passive or active devices
- ➤ Used in receivers, transmitters and instrumentation
- ➤ Superheterodyne receivers combines RF and LO signals to produce IF



Nonlinear Behavior

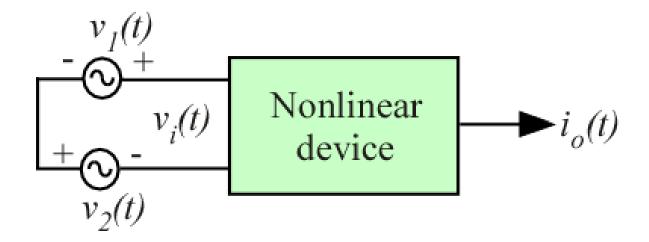


$$i_o(t) = I_0 + av_i(t) + b[v_i(t)]^2 + c[v_i(t)]^3 + \dots$$

- If input has one frequency, harmonics will be generated
- If several input frequencies are used, sum and difference frequencies are generated



Nonlinear Behavior



Square-law behavior is ideal for mixer since the least number of undesired frequencies are produced

$$i_o(t) = av_i(t) + b \left[v_i(t) \right]^2$$



Mixer Principles

$$v_i(t) = V_1 \cos \omega_1 t + V_2 \cos \omega_2 t$$

$$i_o(t) = aV_1 \cos \omega_1 t + aV_2 \cos \omega_2 t + bV_1^2 \cos^2 \omega_1 t$$
$$+ bV_2^2 \cos^2 \omega_2 t + 2bV_1V_2 \cos \omega_1 t \cos \omega_2 t$$

- The final term (or *product term*) yields the desired output

$$2bV_1V_2\cos\omega_1t\cos\omega_2t = bV_1V_2\left[\cos\left(\omega_1-\omega_2\right)t+\cos\left(\omega_1+\omega_2\right)t\right]$$



Mixer Principles

If filter is used to remove unwanted terms:

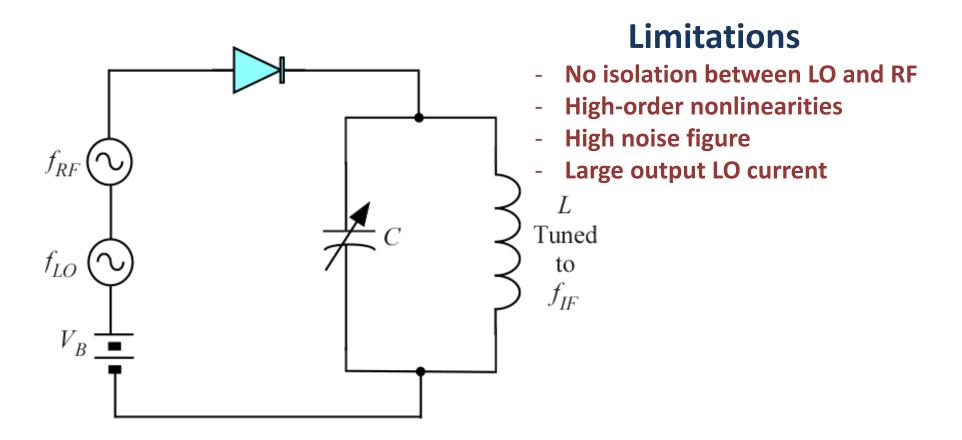
 $\omega_1 \rightarrow \omega_{lo}$: local oscillator frequency

 $\omega_2 \rightarrow \omega_{RF}$: RF frequency

$$i_{o}(t) = bV_{1}V_{2} \left[\cos(\omega_{LO} - \omega_{RF})t + \cos(\omega_{LO} + \omega_{RF})t \right]$$

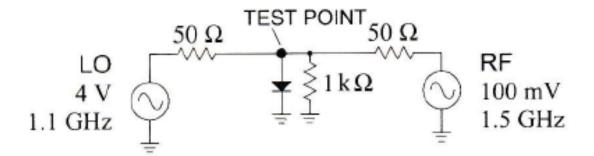
Filter can be used to choose IF frequency

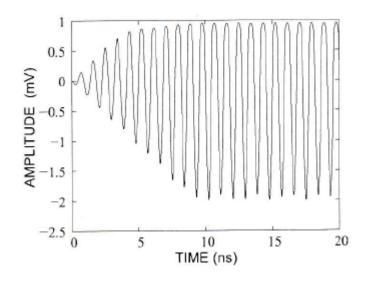
Single-Ended Mixer





Single-Ended Mixer

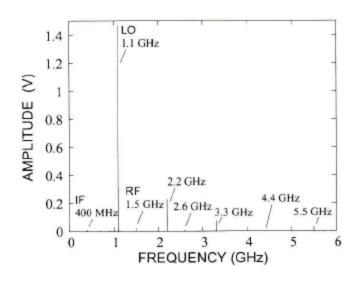


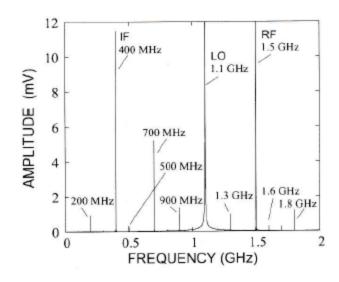


Waveform at test point



Single-Ended Mixer





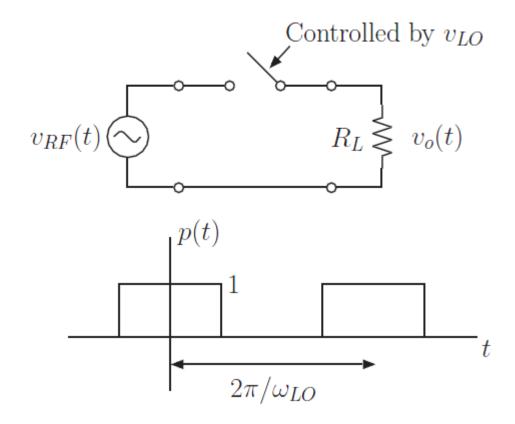
Spectrum at test point

Expanded spectrum at test point

Desired signal is IF at 400 MHz



Switching Principle



Simple switch exhibits strong nonlinearities



Switching Principle

$$S(t) = \frac{1}{2} + \sum_{n=1}^{\infty} \frac{\sin n\pi / 2}{n\pi / 2} \cos n\omega_{LO}t$$

$$v_{RF}(t) = V_{RF} \cos \omega_{RF} t$$

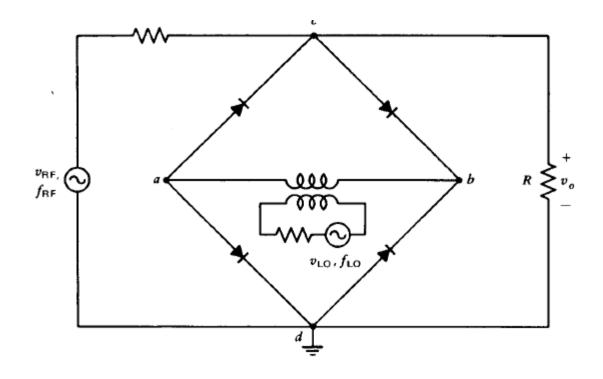
$$v_o(t) = v_{RF}(t) \times S(t) = V_{RF} \cos \omega_{RF} t \left(\frac{1}{2} + \sum_{n=1}^{\infty} \frac{\sin n\pi / 2}{n\pi / 2} \cos n\omega_{LO} t \right)$$

- \triangleright Even harmonics of f_{LO} disappear since $sin(n\pi/2)/n\pi/2 = 0$ when n is even
- \triangleright No frequency terms in f_{LO} and its odd harmonics
- \triangleright Only terms due to $\cos \omega_{RF} t \times \cos n \omega_{LO} t$ with n odd

Filter can be used to select $\left|f_{\mathit{RF}} \pm f_{\mathit{LO}}\right|$



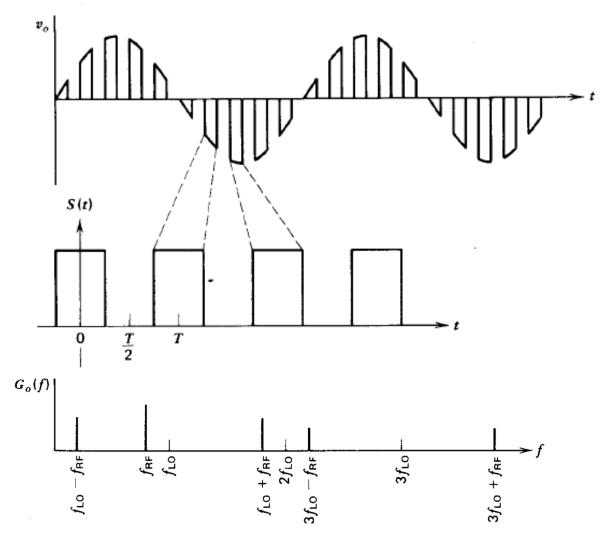
Single-Balanced Diode Mixer



- $\gt{v_{LO}}\gt\gt{v_{RF}}$ so LO signal controls diode states
- **▶** Diodes act like switches

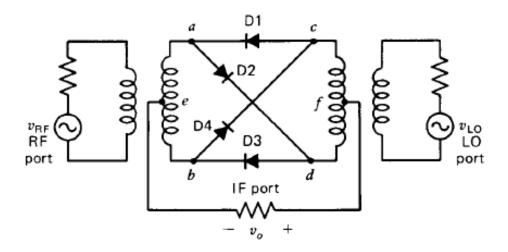


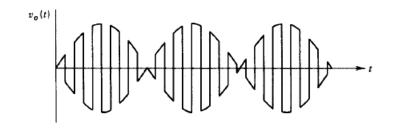
Single-Balanced Diode Mixer Operation





Double-Balanced Diode Mixer



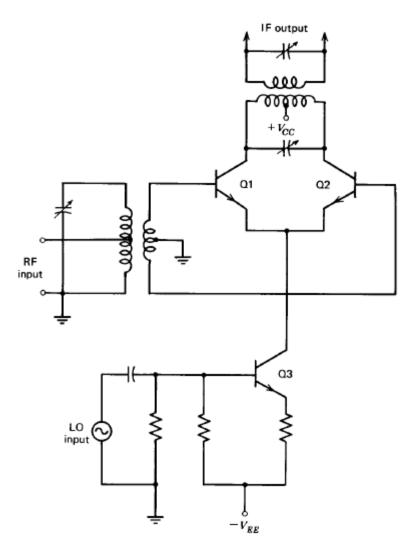


 $\begin{array}{c|c} S(t) \\ \hline \\ 0 \\ \hline \\ \hline \\ T \\ \hline \end{array}$

Currents due to v_{RF} are small compared to those due to v_{LO} .



Balanced BJT Mixer



If transistors are matched, LO signal does not reach RF or IF ports.

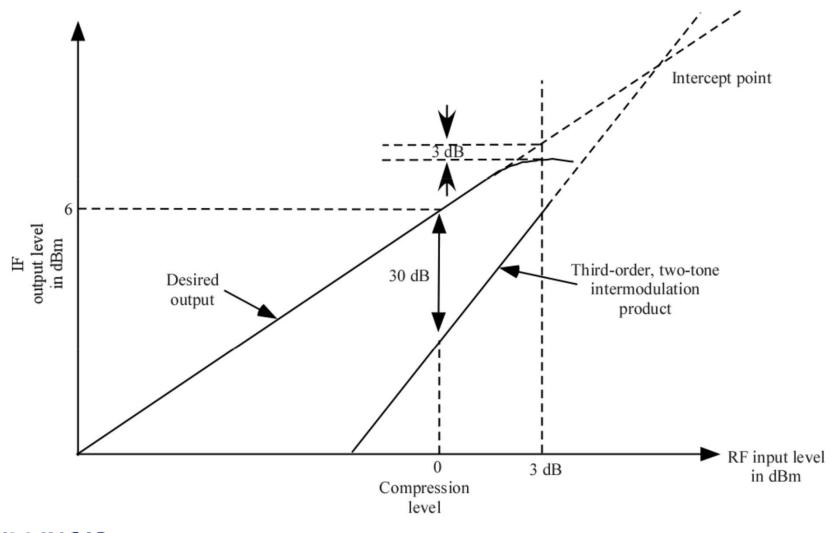


Mixer Performance

- **➤** Conversion Gain (Loss) ratio of IF power to RF power
- \triangleright Noise Figure ratio of SNR_{in} to SNR_{out}
- ➤ Isolation amount of leakage or feedthrough between mixer ports
- Conversion Compression deviation of IF output power from linearity
- > Dynamic Range amplitude range over which mixer can operate
- ➤ Two-Tone, Third-Order Intermodulation Distortion distortion caused by second received RF signal
- ➤ Intercept Point measures third-order suppression
- Desensitization compression of signal at adjacent frequency
- ➤ Harmonic Intermodulation Distortion mixing of harmonics
- Cross-Modulation Distortion transfer from modulated to unmodulated carrier



Mixer Performance Terminology





Mixer Topologies

Mixer type	Advantages	Disadvantages
Active (vs. passive)	Conversion gain Better linearity Lower LO power Simpler to implement	Typically higher noise figure Less-predictable performance Limited to lower frequencies
Unbalanced (active)	Lowest noise figure Ali ports single-ended	Poor port-to-port isolation Poor linearity Difficult to implement
Single-balanced (active)	LO-to-RF isolation RF-to-IF isolation Best linearity Good noise figure	Differential IF output LO-to-IF feed-through
Double-balanced (active)	LO-to-RF, LO-to-IF and RF-to-IF isolation Good spurious product rejection Good linearity Simple to implement	High noise figure High power consumption

